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HAL Id: inria-00536917

https://inria.hal.science/inria-00536917v2

Submitted on 29 Nov 2010 (v2), last revised 28 Sep 2011 (v3)

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N° 7453

Novembre 2010

_ Audio, Speech, and Language Processing _

*R*apport *de recherche*

INRIA

centre de recherche
RENNES – BRETAGNE ATLANTIQUE

A General Flexible Framework for the Handling of Prior Information in Audio Source Separation

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Abstract: Most of audio source separation methods are developed for a particular scenario characterized by the number of sources and channels and the characteristics of the sources and the mixing process. In this paper we introduce a general audio source separation framework based on a library of structured source models that enable the incorporation of prior knowledge about each source via user-specifiable constraints. While this framework generalizes several existing audio source separation methods, it also allows to imagine and implement new efficient methods that were not yet reported in the literature. We first introduce the framework by describing the model structure and constraints, explaining its generality, and summarizing its algorithmic implementation using a generalized expectation-maximization algorithm. Finally, we illustrate the above-mentioned capabilities of the framework by applying it in several new and existing configurations to different source separation problems.

Key-words: Audio source separation, local Gaussian model, nonnegative matrix factorization, expectation-maximization

Un Cadre Général et Flexible pour l'Exploitation de l'Information A Priori en Séparation de Sources Audio

Résumé : La plupart des méthodes de séparation de sources audio sont développées pour un scénario particulier caractérisé par le nombre de sources et de capteurs, ainsi que par les propriétés des sources et du processus de mélange. Dans cet article, nous introduisons un cadre général pour la séparation de sources audio basé sur une librairie de modèles de sources structurés permettant l'exploitation des connaissances a priori sur chaque source à l'aide de contraintes spécifiées par l'utilisateur. Ce cadre généralise plusieurs méthodes de séparation de sources existantes et permet également d'imaginer et d'implémenter de nouvelles méthodes non étudiées dans la littérature. Nous introduisons le cadre proposé en décrivant la structure du modèle et des contraintes, en expliquant sa généralité, et en décrivant son implémentation par un algorithme de maximisation de l'espérance généralisé. Puis nous illustrons le potentiel de cette approche en l'appliquant à plusieurs problèmes de séparation de sources dans des configurations existantes et nouvelles.

Mots-clés : Séparation de sources audio, modèle gaussien local, factorisation matricielle positive, maximisation de l'espérance
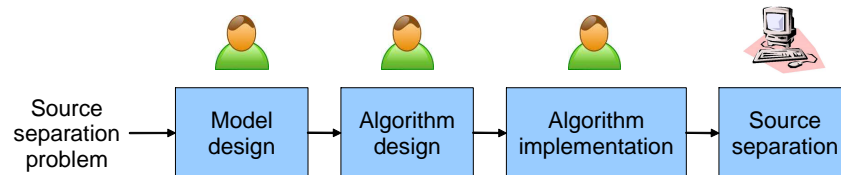
1 Introduction

Separating audio sources from multichannel mixtures is still challenging in most situations. The main difficulty is that audio source separation problems are usually mathematically ill-posed and to succeed one needs to incorporate additional knowledge about the mixing process and/or the source signals. Thus, efficient source separation methods are usually developed for a particular source separation problem characterized by a certain *problem dimensionality*, e.g., determined or underdetermined, certain *mixing process characteristics*, e.g., instantaneous or convolutive, and certain *source characteristics*, e.g., speech, singing voice, drums, bass or noise [1]. For example, a source separation problem may be formulated as follows:

> *"Separate three static sources recorded by two microphones in a reverberant environment, given that two sources are played by percussive musical instruments, the direction of arrival of the first source is known, and the third source is female speech."*

Given a source separation problem, one typically must introduce as much knowledge about this problem as possible into the corresponding separation method so as to achieve good separation performance. However, there is often no common formulation describing methods applied for different problems, and this makes it difficult to reuse a method for a problem it was not originally conceived for. Thus, given a new source separation problem, the common approach consists in (*i*) model design, taking into account problem formulation, (*ii*) algorithm design and (*iii*) implementation (see Fig. 1, top).

Figure 1: Current way of addressing a new source separation problem (top) and the way of addressing it using the proposed flexible framework (bottom).

The motivation of this work is to improve over this time-consuming process by designing a general audio source separation framework that can be applied to virtually any source separation problem by simply selecting from a library of constraints suitable constraints accounting for the available information about that source (see Fig. 1, bottom). More precisely, we wish such a framework to be

- *general*, i.e., generalizing existing methods and making it possible to combine them,

- *flexible*, allowing easy incorporation of the *a priori* knowledge about a particular problem considered.

To achieve the property of generality, we need to find some common formulation for methods we would like to generalize. Many recently proposed methods for audio source separation and/or characterization [2, 3, 4, 5, 6, 7, 8, 9, 10, 11, 12, 13, 14, 15, 16, 17, 18, 19] (see also [1] and references therein) are based on the same so-called *local Gaussian model* describing both the properties of the sources and of the mixing process. Thus, we chose this model as the core of our framework. To achieve flexibility, we fix the global structure of Gaussian covariances, and by means of a parametric model allow the introduction of knowledge about each individual source and its mixing characteristics via constraints on individual parameter subsets. We implement our framework using a generalized expectation-maximization (GEM) algorithm [20], where the M-step is solved by alternating optimization of different parameter subsets, taking the corresponding constraints into account. Such an implementation is in fact possible thanks to the Gaussianity assumption leading to closed form update equations. In summary, our framework generalizes the methods from [2, 3, 4, 5, 6, 7, 8, 9, 10, 11, 12, 13, 14, 15, 16, 17, 18, 19], and, thanks to its flexibility, it becomes applicable in many other scenarios one can imagine.

Our approach is in line with the *library of components* by Cardoso *et al* [21] developed for the separation of components in astrophysical images. However, we consider advanced audio-specific structures inspired by [22,1] for source spectral power, as opposed to the unique block structure in [21] based on the assumption that source power is constant in some pre-defined region of time and space. In that sense, our framework is more flexible than [21]. Besides the framework itself, we propose a new structure for nonnegative matrix factorization (NMF)-like decompositions of source power spectrograms, where the temporal envelope associated with each spectral pattern is represented as a nonnegative linear combination of time-localized temporal patterns. This structure can be used to ensure temporal continuity, but also to model more complex temporal characteristics, such as the attack or decay parts of a note. As compared to [23], where some preliminary aspects of this work were presented, we here present the framework in details, describe its implementation, and extend the experimental part illustrating the framework. Moreover, we propose an original mixing model formulation that allows the representation and the estimation of rank-1 [5] and full-rank [19] (actually any rank) spatial mixing models in a homogeneous way, thus enabling the combination of both models within a given mixture. Finally, we provide a proper probabilistic formulation of local Gaussian modeling for quadratic time-frequency representations [18] that supports and justifies the formulation given in [18].

The rest of this paper is organized as follows. In section 2, existing approaches generalized by the proposed framework are discussed and an overview of the framework is given. Sections 3 and 4 provide a detailed description of the framework and its algorithmic implementation. The results of a few source separation experiments are given in section 5 to illustrate the flexibility of our framework and its potential performance improvement compared to individual approaches. Conclusions are drawn in section 6.

2 Related existing approaches and framework overview

Source separation methods based on the local Gaussian model can be characterized by the following assumptions [2, 5, 13, 19, 1]:

1. *Gaussianity:* in some time-frequency (TF) representation the sources are modeled in each TF bin by zero-mean Gaussian random variables.

2. *Independence:* conditionally to their covariance matrices, these random variables are independent over time, frequency and between sources.

3. *Factorization of spectral and spatial characteristics:* for each TF bin, the covariance matrix of each source is expressed as the product of a *spatial covariance* matrix representing its spatial characteristics and a scalar *spectral power* representing its spectral characteristics.

4. *Linearity of mixing:* the mixing process translates into addition in the covariance domain.

2.1 State-of-the-art approaches based on the local Gaussian model

The state-of-the-art approaches [2,3,4,5,6,7,8,9,10,11,12,13,14,15,16,17,18,19] cover a wide range of source separation problems and models expressed via particular structures of local Gaussian covariances, including:

1. *Problem dimensionality:* Denoting by I and J, respectively, the number of channels of the observed mixture and the number of sources to separate, the *single-channel* ($I = 1$) case is addressed in [6], and *underdetermined* ($1 < I < J$) and *(over-)determined* ($I \geq J$) cases are addressed in [5] and [2], respectively.

2. *Spatial covariance model: Instantaneous* and *convolutive* mixtures of point sources are modeled by *rank-1* spatial covariance matrices in [5] and [3], respectively, and highly reverberant mixtures of point sources are modeled by *full-rank* spatial covariance matrices in [19].

3. *Spectral power model:* Several models were proposed for the spectral power, e.g., *unconstrained* models [10], *block constant* models [5], Gaussian mixture models (GMM) or hidden Markov models (HMM) [2], Gaussian scaled mixture models (GSMM) or scaled HMMs (S-HMM) [13], nonnegative matrix factorization (NMF) [4] together with its variants, harmonic NMF [14] or temporal activation constrained NMF [9], and source-filter models [16].

These models are suitable for the representation of different types of sources, for example GSMM is rather suitable for a monophonic source, e.g., speech, and NMF for a polyphonic one, e.g., polyphonic musical instrument, [13].

4. *Input representation:* While the most of the considered methods use the short time Fourier transform (STFT) as the input TF representation, some of them, e.g., [14,15,18], use the auditory-motivated equivalent rectangular bandwidth (ERB) quadratic representation. More generally, we consider here both *linear representations*, where the signal is represented by a vector of complex-valued coefficients in each TF bin, as well as *quadratic representations*, where the signal is represented via its local covariance matrix in each TF bin [24].

Table 1 provides an overview of some of the local Gaussian model-based approaches considered here, where the specificities of each method are marked by crosses ×. We see from table 1 that a few of these methods have already been combined together, for example GSMM and NMF were combined in [8], and NMF [9] was combined with rank-1 and full-rank mixing models in [13] and [17], respectively. However, many combinations have not yet been investigated.

	Reference	[7]	[6]	[8]	[16]	[4]	[14]	[15]	[9]	[5]	[11]	[13]	[19]	[18]	[17]	[3]	[2]
Problem dimensionality	single-channel	×	×	×	×	×	×	×	×								
	underdetermined									×	×	×	×	×	×		
	(over-)determined															×	×
Spatial covariance model	rank-1 instantaneous									×	×						
	rank-1 convolutive											×				×	×
	full-rank												×	×	×		
Spectral power model	unconstrained												×	×	×		
	block constant									×							
	GMM / HMM	×									×						
	GSMM / S-HMM		×	×	×												
	NMF			×	×	×						×					
	harmonic NMF						×	×									
	temp. constr. NMF							×	×								
	source-filter	×			×												
Input representation	linear	×	×	×	×	×			×	×	×	×	×		×	×	×
	quadratic						×	×						×			

Table 1: Some state-of-the-art local Gaussian model-based approaches for audio source separation.

2.2 Other related state-of-the-art approaches

While the local Gaussian model-based framework offers maximum of flexibility, there exist some methods that do not satisfy (fully or partially) the aforementioned assumptions and are thus not strictly covered by the framework. Nevertheless, our framework allows the implementation of similar structures. Let us give some examples. Binary masking-based source estimation [25, 26] does not satisfy the source independence assumption. However, it is known to perform poorly compared to local Gaussian model-based separation, as it was shown in [13, 18] for convolutive mixtures [1] and demonstrated through the signal separation evaluation campaigns SiSEC 2008 [28] and SiSEC 2010 [27], where for instantaneous mixtures local Gaussian model-based approaches gave better results than the *oracle* (using the ground truth) binary masks. The methods proposed in [29, 30] are also based on Gaussian models albeit in the time domain. Notably, time sample-based GMMs and time-varying autoregressive models are considered as source models in [29] and [30], respectively. However, the number of existing time-domain structures is fairly reduced. Our TF domain models make it possible to account for these structures by means of suitable constraints over spectral power, while allowing their combination with more advanced structures. There are also many works on NMF and its extensions [31, 32, 33, 34, 35, 36] and on GMMs / HMMs [37, 38] based on nongaussian models of the complex-valued STFT coefficients. These models are essentially covered by our framework in the sense that we can implement similar or equivalent model structures, albeit under Gaussian assumptions. The benefit of local Gaussian modeling is that it naturally leads to closed-form expressions in the multichannel case, contrary to the models in [31, 32, 33, 34, 35, 36, 37, 38].

2.3 Framework overview

We now present an overview of the proposed framework focusing on the most important concepts. An exhaustive description is given in sections 3 and 4.

The framework is based on a flexible model described by parameters $\theta = \{\theta_j\}_{j=1}^{J}$, where θ_j are the parameters of the j-th source ($j = 1, \ldots, J$). Each θ_j is split in turn into nine parameter subsets according to a fixed structure, as described below and summarized in table 2.

2.3.1 Model structure

The parameters of j-th source include a complex-valued tensor $\mathbf{A}_j$ modeling its spatial covariance, and eight nonnegative matrices $(\theta_{j,2}, \ldots, \theta_{j,9})$ modeling its spectral power over all TF bins.

The spectral power, denoted as $\mathbf{V}_j$, is assumed to be the product of an *excitation spectral power* $\mathbf{V}_j^{\mathrm{ex}}$, representing, e.g., the excitation of the glottal source for voice or the plucking of the string of a guitar, and a *filter spectral power* $\mathbf{V}_j^{\mathrm{ft}}$, representing, e.g., the vocal tract or the impedance of the guitar body [33, 22]. While such a model is usually called source-filter model, we call

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[1] Binary masking-based approaches can still be quite powerful for convolutive mixtures, as demonstrated in [27]. Thus, a good way to proceed is probably to use them to initialize local Gaussian model-based approaches, as it is done in [13], and as we do in the experimental part.

it here *excitation-filter model* in order to avoid possible confusions with the "sources" to be separated.

The excitation spectral power $\mathbf{V}_j^{\mathrm{ex}}$ is further decomposed as the sum of *characteristic spectral patterns* $\mathbf{E}_j^{\mathrm{ex}}$ modulated by *time activation coefficients* $\mathbf{P}_j^{\mathrm{ex}}$ [4, 9]. Each characteristic spectral pattern may be associated for instance with one specific pitch, so that the time activation coefficients denote which pitches are active on each time frame. In order to further constrain the fine structure of the spectral patterns, they are represented as linear combinations of *narrowband spectral patterns* $\mathbf{W}_j^{\mathrm{ex}}$ [14] with weights $\mathbf{U}_j^{\mathrm{ex}}$. These narrowband patterns may be for instance harmonic, inharmonic or noise-like and the weights determine the overall spectral envelope. Following the same idea, we propose here to represent the series of time activation coefficients $\mathbf{P}_j^{\mathrm{ex}}$ as sums of *time-localized patterns* $\mathbf{H}_j^{\mathrm{ex}}$ with weights $\mathbf{G}_j^{\mathrm{ex}}$. The time-localized patterns may represent the typical temporal shape of the notes while the weights encode their onset times. Different temporal fine structures such as continuity or specific rhythm patterns may also be accounted for in this way. Note that temporal models of the activation coefficients have been proposed in the state-of-the-art, using probabilistic priors [32, 9], note-specific Gaussian-shaped time-localized patterns [39], or unstructured TF patterns [31]. Our proposition is complementary to [32, 9] in that it accounts for temporal behaviour in the model structure itself in addition to possible priors on the model parameters. Moreover, it is more flexible than [32, 9, 39], since it allows the modeling of other characteristics than continuity or sparsity. Finally, while it can model similar TF patterns to [31], it involves much fewer parameters, which typically leads to more robust parameter estimation.

The filter spectral power $\mathbf{V}_j^{\mathrm{ft}}$ is similarly expressed in terms of characteristic spectral patterns $\mathbf{E}_j^{\mathrm{ft}}$ modulated by time activation coefficients [16], which are in turn decomposed into narrowband spectral patterns $\mathbf{W}_j^{\mathrm{ft}}$ with weights $\mathbf{U}_j^{\mathrm{ft}}$ and time-localized patterns $\mathbf{H}_j^{\mathrm{ft}}$ with weights $\mathbf{G}_j^{\mathrm{ft}}$, respectively. In the case of speech or singing voice, each characteristic spectral pattern may represent the spectral formants of a given phoneme, while the plosiveness and the sequence of pronounced phonemes may be encoded by the time-localized patterns and the associated weights.

In summary, as it will be explained in details in section 3, the spectral power of each source obeys the following hierarchical nonnegative matrix decomposition structure:

$$\mathbf{V}_j = \mathbf{V}_j^{\mathrm{ex}} \odot \mathbf{V}_j^{\mathrm{ft}}, \tag{1}$$

$$\mathbf{V}_j = \left(\mathbf{E}_j^{\mathrm{ex}} \, \mathbf{P}_j^{\mathrm{ex}} \right) \odot \left(\mathbf{E}_j^{\mathrm{ft}} \, \mathbf{P}_j^{\mathrm{ft}} \right), \tag{2}$$

$$\mathbf{V}_j = \left(\mathbf{W}_j^{\mathrm{ex}} \, \mathbf{U}_j^{\mathrm{ex}} \, \mathbf{G}_j^{\mathrm{ex}} \, \mathbf{H}_j^{\mathrm{ex}} \right) \odot \left(\mathbf{W}_j^{\mathrm{ft}} \, \mathbf{U}_j^{\mathrm{ft}} \, \mathbf{G}_j^{\mathrm{ft}} \, \mathbf{H}_j^{\mathrm{ft}} \right), \tag{3}$$

where $\odot$ denotes element-wise matrix multiplication.

Figure 2 gives an example of the excitation-filter decomposition (1) as applied to the spectral power of several guitar notes. In this example the filter $\mathbf{V}_j^{\mathrm{ft}}$ is time-invariant with lowpass characteristics, and the excitation $\mathbf{V}_j^{\mathrm{ex}}$ is a time-varying combination of few characteristic spectral patterns.

Figure 3 shows an example of the excitation structure $\mathbf{V}_j^{\mathrm{ex}} = \mathbf{W}_j^{\mathrm{ex}} \, \mathbf{U}_j^{\mathrm{ex}} \, \mathbf{G}_j^{\mathrm{ex}} \, \mathbf{H}_j^{\mathrm{ex}}$, as applied to six notes played on a xylophone. In this example, the narrowband spectral patterns $\mathbf{W}_j^{\mathrm{ex}}$ include 66 harmonic patterns modeling the harmonic

part of 11 notes and 9 smooth patterns modeling the attacks, and the matrix of weights $\mathbf{U}_j^{\mathrm{ex}}$ is very sparse so as to eliminate invalid combinations of narrowband spectral patterns (e.g., a characteristic spectral pattern should not be a combination of narrowband spectral patterns with different pitches). The time-localized patterns $\mathbf{H}_j^{\mathrm{ex}}$ include decreasing exponentials to model the decay part of the notes and discrete Dirac functions to model note attacks, and the matrix of weights $\mathbf{G}_j^{\mathrm{ex}}$ is sparse so as not to allow the attacks (smooth spectral patterns) to be modulated by exponential temporal patterns and not to allow harmonic note parts (harmonic spectral patterns) to be modulated by Dirac temporal patterns.

Parameter subsets		Size	Range
$\theta_{j,1} = \mathbf{A}_j$	mixing parameters	$I \times R_j \times F \times N$	$\in \mathbb{C}$
$\theta_{j,2} = \mathbf{W}_j^{\mathrm{ex}}$	ex. narrowband spectral patterns	$F \times L_j^{\mathrm{ex}}$	$\in \mathbb{R}_+$
$\theta_{j,3} = \mathbf{U}_j^{\mathrm{ex}}$	ex. spectral pattern weights	$L_j^{\mathrm{ex}} \times K_j^{\mathrm{ex}}$	$\in \mathbb{R}_+$
$\theta_{j,4} = \mathbf{G}_j^{\mathrm{ex}}$	ex. time pattern weights	$K_j^{\mathrm{ex}} \times M_j^{\mathrm{ex}}$	$\in \mathbb{R}_+$
$\theta_{j,5} = \mathbf{H}_j^{\mathrm{ex}}$	ex. time-localized patterns	$M_j^{\mathrm{ex}} \times N$	$\in \mathbb{R}_+$
$\theta_{j,6} = \mathbf{W}_j^{\mathrm{ft}}$	ft. narrowband spectral patterns	$F \times L_j^{\mathrm{ft}}$	$\in \mathbb{R}_+$
$\theta_{j,7} = \mathbf{U}_j^{\mathrm{ft}}$	ft. spectral pattern weights	$L_j^{\mathrm{ft}} \times K_j^{\mathrm{ft}}$	$\in \mathbb{R}_+$
$\theta_{j,8} = \mathbf{G}_j^{\mathrm{ft}}$	ft. time pattern weights	$K_j^{\mathrm{ft}} \times M_j^{\mathrm{ft}}$	$\in \mathbb{R}_+$
$\theta_{j,9} = \mathbf{H}_j^{\mathrm{ft}}$	ft. time-localized patterns	$M_j^{\mathrm{ft}} \times N$	$\in \mathbb{R}_+$

Table 2: Parameter subsets $\theta_{j,k}$ $(j = 1, \ldots, J, \ k = 1, \ldots, 9)$ encoding the structure of each source.
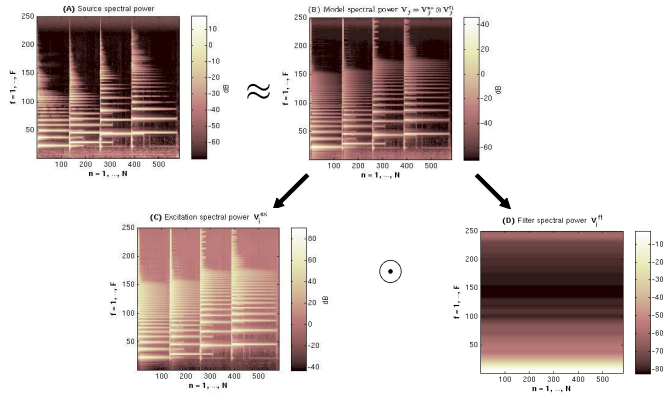

Figure 2: Excitation-filter decomposition as applied to the spectral power of several guitar notes. (**A**): source spectral power, (**B**): model spectral power $\mathbf{V}_j = \mathbf{V}_j^{\mathrm{ex}} \odot \mathbf{V}_j^{\mathrm{ft}}$, (**C**): excitation spectral power $\mathbf{V}_j^{\mathrm{ex}}$, (**D**): filter spectral power $\mathbf{V}_j^{\mathrm{ft}}$.
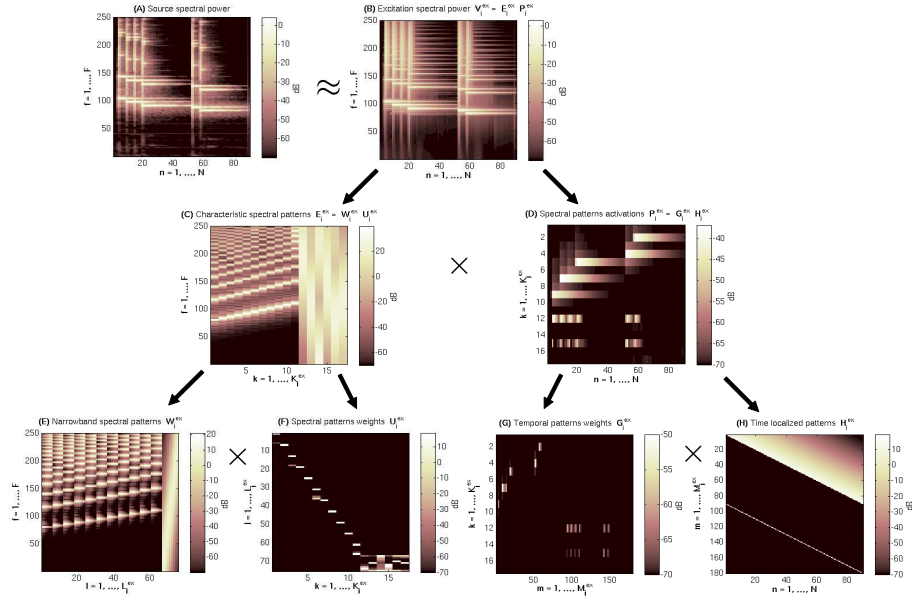
Figure 3: Excitation power decomposition $\mathbf{V}_j^{\text{ex}} = \mathbf{W}_j^{\text{ex}} \mathbf{U}_j^{\text{ex}} \mathbf{G}_j^{\text{ex}} \mathbf{H}_j^{\text{ex}}$ as applied to the spectral power of several xylophone notes. (**A**): source spectral power, (**B**): excitation spectral power $\mathbf{V}_j^{\text{ex}} = \mathbf{E}_j^{\text{ex}} \mathbf{P}_j^{\text{ex}}$, (**C**): characteristic spectral patterns $\mathbf{E}_j^{\text{ex}} = \mathbf{W}_j^{\text{ex}} \mathbf{U}_j^{\text{ex}}$, (**D**): spectral pattern activations $\mathbf{P}_j^{\text{ex}} = \mathbf{G}_j^{\text{ex}} \mathbf{H}_j^{\text{ex}}$, (**E**): narrowband spectral patterns $\mathbf{W}_j^{\text{ex}}$, (**F**): spectral pattern weights $\mathbf{U}_j^{\text{ex}}$, (**G**): temporal pattern weights $\mathbf{G}_j^{\text{ex}}$, (**H**): time-localized patterns $\mathbf{H}_j^{\text{ex}}$.

2.3.2 Constraints

Given the above fixed model structure, prior information about each source can now be exploited by specifying deterministic or probabilistic constraints over each parameter subset of Table 2. Examples of such constraints are given in Table 3. Each parameter subset can be fixed [2] (i.e., unchanged during estimation), adaptive (i.e., fully fitted to the mixture) or partially adaptive (only some parameters within the subset are adaptive). In the latter two cases, a probabilistic prior, such as a continuity prior [9] or a sparsity-inducing prior [4], can be specified over the parameters. The mixing parameters $\mathbf{A}_j$ can be time-varying or time-invariant (in Table 3 the latter case is only considered), frequency-dependent for convolutive mixtures or frequency-independent for instantaneous mixtures. Mixing parameters $\mathbf{A}_j$ can be given a probabilistic prior as well. E.g., it can be a Gaussian prior with the mean corresponding to the parameters of a presumed direction and with the covariance matrix representing a degree of uncertainty about this direction. The rank R_j $(1 \leq R_j \leq I)$ of the spatial covariance is specifiable via the size of tensor $\mathbf{A}_j$ (see Table 2). Each parameter subset may also be constrained to have a limited number of nonzero entries. For instance, every column of $\mathbf{G}_j^{\mathrm{ex}}$ and / or $\mathbf{G}_j^{\mathrm{ft}}$ may be constrained to have a single nonzero entry accounting for a GSMM / S-HMM structure or a single nonzero entry equal to 1 accounting for a GMM / HMM structure.

For example, the following constraints can be chosen for the source separation problem mentioned at the beginning of the introduction. The mixing parameters $\mathbf{A}_j$ of each source are convolutive and full-rank $(R_j = 2)$, since the recording environment is convolutive and reverberant, and they are time-invariant, since the sources are static. Parameters $\mathbf{A}_1$ are given a Gaussian prior using the information about the first source direction of arrival. The spectral powers of the first and second source (percussive musical instruments) can be constrained to $\mathbf{V}_j = \mathbf{W}_j^{\mathrm{ex}} \mathbf{G}_j^{\mathrm{ex}} \mathbf{H}_j^{\mathrm{ex}}$ [3] $(j = 1, 2)$ with fixed spectral patterns $\mathbf{W}_j^{\mathrm{ex}}$ pre-trained from some database of percussive sounds, adaptive time pattern weights $\mathbf{G}_j^{\mathrm{ex}}$, and fixed manually constructed time-localized patterns $\mathbf{H}_j^{\mathrm{ex}}$ representing few characteristic time-localized patterns of percussive sounds. The spectral power of the third source (female speech) can be constrained to $\mathbf{V}_3 = \mathbf{W}_3^{\mathrm{ex}} \mathbf{U}_3^{\mathrm{ex}} \mathbf{G}_3^{\mathrm{ex}} \mathbf{H}_3^{\mathrm{ex}}$ [3] with fixed manually constructed harmonic an smooth narrowband spectral patterns (see, e.g., Fig. 3 (E)), adaptive spectral and time pattern weights $\mathbf{U}_3^{\mathrm{ex}}$ and $\mathbf{G}_3^{\mathrm{ex}}$, and fixed time-localized patterns $\mathbf{H}_3^{\mathrm{ex}}$ consisting of some Gaussian-shaped patterns to model speech spectrum continuity. The time pattern weights $\mathbf{G}_3^{\mathrm{ex}}$ could be also constrained to a GSMM, to model speech monophonicity.

2.3.3 Estimation algorithm

Given the above model structure and constraints, source separation can be achieved in two steps as shown in Fig. 4. First, given initial parameter values,

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[2] The fixed parameters can be either set manually or learned beforehand from some training data. Learning is equivalent to model parameter estimation over the training data and can thus be achieved using our framework.

[3] Note that any set of matrices can be virtually removed from the spectral power decomposition (3). For example, one can obtain $\mathbf{V}_j = \mathbf{W}_j^{\mathrm{ex}} \mathbf{G}_j^{\mathrm{ex}} \mathbf{H}_j^{\mathrm{ex}}$ by assuming that the matrices $\mathbf{W}_j^{\mathrm{ft}}, \mathbf{U}_j^{\mathrm{ft}}, \mathbf{G}_j^{\mathrm{ft}}$ and $\mathbf{H}_j^{\mathrm{ft}}$ are of sizes $F \times 1$, 1×1, 1×1, and $1 \times N$, and that all their entries are fixed to 1, and that $\mathbf{U}_j^{\mathrm{ex}} = \mathbf{I}_{K_j^{\mathrm{ex}}}$ is the $K_j^{\mathrm{ex}} \times K_j^{\mathrm{ex}}$ identity matrix.

Parameter subsets	Constraint	Value

$\mathbf{A}_j, \mathbf{W}_j^{\text{ex}}, \mathbf{U}_j^{\text{ex}}, \mathbf{G}_j^{\text{ex}}, \mathbf{H}_j^{\text{ex}}, \mathbf{W}_j^{\text{ft}}, \mathbf{U}_j^{\text{ft}}, \mathbf{G}_j^{\text{ft}}, \mathbf{H}_j^{\text{ft}}$	degree of adaptability	`'fixed'`
		 `'part_adapt'`
		 `'adapt'`

$\mathbf{A}_j$	mixing stationarity	`'time_inv'`
	mixing type	`'conv'` `'inst'`

$\mathbf{G}_j^{\text{ex}}, \mathbf{G}_j^{\text{ft}}$	temporal constraint	`'null'`
		 `'GMM'`, `'HMM'`
		 `'GSMM'`, `'SHMM'`

Table 3: Examples of user-specifiable constraints over the parameter subsets.

the model parameters θ are estimated from the mixture $\mathbf{X}$ using an iterative GEM algorithm, where the E-step consists in computing some quantity $\widehat{\mathbb{T}}$ called *conditional expectation of the natural statistics*, and the M-step consists in updating the parameters θ given $\widehat{\mathbb{T}}$ by alternating optimization of each of the $J \times 9$ parameter subsets. This allows taking any combination of constraints specified by user into account. Second, given the mixture $\mathbf{X}$ and the estimated model parameters θ, source estimates $\widehat{\mathbf{Y}}$ are computed using Wiener filtering.
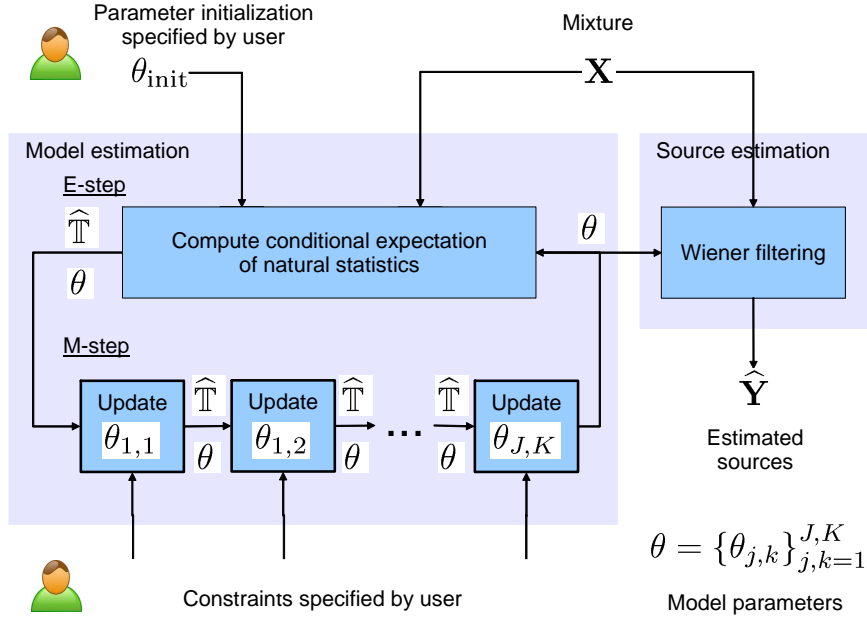

Figure 4: Overview of the proposed general algorithm for parameter estimation and source separation.

2.4 Generality

It can be easily shown that the model structures considered in $[2,3,4,5,6,7,8,9,$ $10,11,12,13,14,15,16,17,18,19]$ are particular instances of the proposed general formulation. Let us give some examples.

Pham *et al* [3] assume rank-1 spatial covariances and constant spectral power over time-frequency regions of size 1 frequency bin $\times$ L frames. This structure can be implemented in our framework by choosing rank-1 adaptive spatial time-invariant covariances, i.e., $\mathbf{A}_j$ is an adaptive tensor of size $2 \times 1 \times F \times N$ subject to time-invariance constraint, and constraining the spectral power to $\mathbf{V}_j = \mathbf{W}_j^{\text{ex}} \mathbf{G}_j^{\text{ex}} \mathbf{H}_j^{\text{ex}}$ [3] with $\mathbf{W}_j^{\text{ex}}$ being the $F \times F$ identity matrix, $\mathbf{G}^{\text{ex}}$ a $F \times \lceil N/L \rceil$ adaptive matrix, and $\mathbf{H}_j^{\text{ex}}$ the $\lceil N/L \rceil \times N$ fixed matrix with entries $h_{j,mn}^{\text{ex}} = 1$ for $n \in \mathcal{L}_m$ and $h_{j,mn}^{\text{ex}} = 0$ for $n \notin \mathcal{L}_m$, where $\mathcal{L}_m$ is the set of time frames belonging to the m-th block.

Multichannel NMF structures with point source (rank-1) [13] or diffuse source (full-rank) [17] models can be represented within our framework as $\mathbf{V}_j = \mathbf{W}_j^{\text{ex}} \mathbf{G}_j^{\text{ex}}$ [3] with $\mathbf{W}_j^{\text{ex}}$ and $\mathbf{G}_j^{\text{ex}}$ being adaptive matrices of size $F \times K_j^{\text{ex}}$ and $K_j^{\text{ex}} \times N$, respectively, and $\mathbf{A}_j$ being an adaptive tensor of size $2 \times 1 \times F \times N$ or $2 \times 2 \times F \times N$, respectively, subject to time-invariance constraint.

Excitation-filter model-based separation of the main melody vs. the background music from single-channel recordings by Durrieu *et al.* [16] can be represented within our framework as follows. Mixing parameters $\mathbf{A}_j$ ($j = 1, 2$) are assumed to form a tensor of size $1 \times 1 \times F \times N$ with all the entries fixed to 1. The background music spectral power $\mathbf{V}_1$ is modeled exactly as in the case of the multichannel NMF described in the previous paragraph. The main melody spectral power is constrained to $\mathbf{V}_2 = (\mathbf{W}_2^{\text{ex}} \mathbf{G}_2^{\text{ex}}) \odot (\mathbf{W}_2^{\text{ft}} \mathbf{G}_2^{\text{ft}})$ [3] with $\mathbf{W}_2^{\text{ex}}$ being fixed and $\mathbf{G}_2^{\text{ex}}$, $\mathbf{W}_2^{\text{ft}}$ and $\mathbf{G}_2^{\text{ft}}$ being adaptive. Without any supplementary constraints this model is equivalent to the model referred as *instantaneous mixture model* in [16], and applying GSMM constraints to both the matrices $\mathbf{G}_2^{\text{ex}}$ and $\mathbf{G}_2^{\text{ft}}$ this model is equivalent to the model referred as *GSMM* in [16].

2.5 Current baseline implementation

We have so far implemented a baseline version of the framework in Matlab that covers only the library of constraints summarized in table 3 for mono or stereo recordings ($I = 1$ or $I = 2$). This restriction to up to $I = 2$ channels enables the use of a 2×2 matrix inversion trick described in [13] that leads to an efficient implementation in Matlab. However, the framework itself is neither restricted to the constraints in table 3 nor to mono / stereo mixtures. We plan to release the code early 2011 following the schedule of the Quaero project.

3 Detailed structure and example constraints

In this section we describe in details the nine parameter subsets modeling each source and some example constraints. We also introduce the detailed notations to be used in the rest of the paper.

3.1 Formulation of the audio source separation problem

We assume that the observed I-channel time-domain signal, called *mixture*, $\tilde{\mathbf{x}}(t) \in \mathbb{R}^I$, $t = 1, \ldots, T$, is the sum of J multichannel signals $\tilde{\mathbf{y}}_j(t) \in \mathbb{R}^I$, called *spatial source images* [1, 21]:

$$\tilde{\mathbf{x}}(t) = \sum\nolimits_{j=1}^{J} \tilde{\mathbf{y}}_j(t). \tag{4}$$

The goal of source separation is to estimate the spatial source images $\tilde{\mathbf{y}}_j(t)$ given the mixture $\tilde{\mathbf{x}}(t)$. This now common formulation is more general than the convolutive formulation in [13], which is restricted to point sources [1, 21].

3.2 Input representation

Audio signals are usually processed in the TF domain, due to their sparsity in this domain. Two families of input representations are considered in the literature, namely *linear* [13] and *quadratic* [18] representations.

3.2.1 Linear representations

After applying a linear complex-valued TF transform, the mixture (4) becomes:

$$\mathbf{x}_{fn} = \sum\nolimits_{j=1}^{J} \mathbf{y}_{j,fn}, \tag{5}$$

where $\mathbf{x}_{fn} \in \mathbb{C}^I$ and $\mathbf{y}_{j,fn} \in \mathbb{C}^I$ are I-dimensional complex-valued vectors of TF coefficients of the corresponding time-domain signals; and $f = 1, \ldots, F$ and $n = 1, \ldots, N$ denote respectively frequency bin and time-frame index. This formulation covers the STFT, that is the most popular TF representation used for audio source separation.

3.2.2 Quadratic representations

A few studies have relied on quadratic representations instead, where the signal is described in each TF bin by its empirical $I \times I$ covariance matrix [5, 10, 18]

$$\widehat{\mathbf{R}}_{\mathbf{x}, fn} = \widehat{\mathbb{E}}[\mathbf{x}_{fn} \mathbf{x}_{fn}^H], \tag{6}$$

where $\widehat{\mathbb{E}}[\cdot]$ denotes *empirical expectation* computed, e.g., by local averaging of the STFT [5, 10] or of the input of an ERB filterbank [18]. Note that linear representations are special cases of quadratic representations with $\widehat{\mathbf{R}}_{\mathbf{x}, fn} = \mathbf{x}_{fn} \mathbf{x}_{fn}^H$. Quadratic representations include additional information about the local correlation between channels which often increases the accuracy of parameters estimation [10]. In the following, we use the linear notations $\mathbf{x}_{fn}$ and $\mathbf{y}_{j,fn}$ for simplicity and include the empirical expectation when appropriate. A more rigorous derivation of the local Gaussian model for quadratic representations is given in Appendix A.

3.3 Local Gaussian model

We assume that in each TF bin, each source $\mathbf{y}_{j,fn} \in \mathbb{C}^I$ is a proper complex-valued Gaussian random vector with zero mean and covariance matrix $\mathbf{\Sigma}_{\mathbf{y},j,fn} = v_{j,fn}\mathbf{R}_{j,fn}$

$$\mathbf{y}_{j,fn} \sim \mathcal{N}_c\left(\bar{0}, v_{j,fn}\mathbf{R}_{j,fn}\right), \tag{7}$$

where the matrix $\mathbf{R}_{j,fn} \in \mathbb{C}^{I \times I}$ called *spatial covariance matrix* represents the spatial characteristics of the source and of the mixing setup, and the non-negative scalar $v_{j,fn} \in \mathbb{R}_+$ called *spectral power* represents the spectral characteristics of the source [1]. Moreover, the random vectors $\mathbf{y}_{j,fn}$ are assumed to be mutually independent given $\mathbf{\Sigma}_{\mathbf{y},j,fn}$.

3.4 Spatial covariance structure and example constraints

3.4.1 Structure

In the case of audio, it is mostly interesting to consider either rank-1 spatial covariances representing instantaneously or convolutively mixed point sources with low reverberation [13] or full-rank spatial covariances modeling diffuse or reverberated sources [19]. More generally, we assume covariances of any positive rank. Let $0 < R_j \leq I$ be the rank of covariance $\mathbf{R}_{j,fn}$. This matrix can then be non-uniquely represented as [4]

$$\mathbf{R}_{j,fn} = \mathbf{A}_{j,fn}\mathbf{A}_{j,fn}^H, \tag{8}$$

where $\mathbf{A}_{j,fn}$ is an $I \times R_j$ complex-valued matrix of rank R_j. Moreover, for every source j and for every TF bin (f,n) we introduce R_j independent Gaussian random variables $s_{jr,fn}$ $(r = 1, \ldots, R_j)$ distributed as

$$s_{jr,fn} \sim \mathcal{N}_c\left(0, v_{j,fn}\right). \tag{9}$$

With these notations the model defined by (5) and (7) is equivalent to the following mixture of $R = \sum_{j=1}^{J} R_j$ point *sub-sources* $s_{jr,fn}$:

$$\mathbf{x}_{fn} = \mathbf{A}_{fn}\mathbf{s}_{fn}, \tag{10}$$

where $\mathbf{s}_{fn} = [\mathbf{s}_{1,fn}^T, \ldots, \mathbf{s}_{J,fn}^T]^T$ is an $R \times 1$ vector of sub-source coefficients with $\mathbf{s}_{j,fn} = [s_{j1,fn}, \ldots, s_{jR_j,fn}]^T$, and $\mathbf{A}_{fn} = [\mathbf{A}_{1,fn}, \ldots, \mathbf{A}_{J,fn}]$ is an $I \times R$ mixing matrix. Thus, for a given TF bin (f,n) our model is equivalent to a complex-valued linear mixture of R sub-sources (10), where the sub-sources $s_{jr,fn}$ $(r = 1, \ldots, R_j)$ associated with the same source j share the same spectral power (9). We suppose that the rank R_j is specified for every source j.

3.4.2 Example constraints

In our baseline implementation implementation we assume that the spatial covariances are time-invariant, i.e., $\mathbf{A}_{j,fn} = \mathbf{A}_{j,f}$. Moreover, we assume that for

---

[4]Such an R_j-rank covariance matrix parametrization was inspired by [21], where $\mathbf{R}_{j,fn}$, intended to model correlated or multi-dimensional components, is parametrized as $\mathbf{R}_{j,fn} = \mathbf{A}_{j,fn}\mathbf{P}_{j,fn}\mathbf{A}_{j,fn}^H$, where $\mathbf{P}_{j,fn}$ is a full-rank $R_j \times R_j$ positive matrix. However, our parametrization (8) is less redundant and it is applied for audio source separation, and not for separation of components in astrophysical images, as in [21].

every source j the spatial parameters $\mathbf{A}_j$ can be either instantaneous (i.e., constant over frequency and real-valued: $\mathbf{A}_{j,fn} = \mathbf{A}_{j,n} \in \mathbb{R}^{I \times R_j}$) or convolutive (i.e., frequency-independent), and either fixed, adaptive or partially adaptive. Some examples of constraints are given in table 3.

3.5 Spectral power structure and example constraints

To model spectral power we use nonnegative hierarchical audio-specific decompositions [22], thus all variables introduced in this section are assumed to be non-negative.

3.5.1 Excitation-filter model

We first model the spectral power $v_{j,fn}$ as the product of an excitation spectral power $v_{j,fn}^{\text{ex}}$ and a filter spectral power $v_{j,fn}^{\text{ft}}$ [33, 22]:

$$v_{j,fn} = v_{j,fn}^{\text{ex}} \times v_{j,fn}^{\text{ft}}, \tag{11}$$

that can be rewritten as (1) introducing matrices $\mathbf{V}_j \triangleq [v_{j,fn}]_{f,n}$, $\mathbf{V}_j^{\text{ex}} \triangleq [v_{j,fn}^{\text{ex}}]_{f,n}$ and $\mathbf{V}_j^{\text{ft}} \triangleq [v_{j,fn}^{\text{ft}}]_{f,n}$.

3.5.2 Excitation power structure

The excitation spectral power $[v_{j,fn}^{\text{ex}}]_f$ is modeled as the sum of K_j^{ex} characteristic spectral patterns $[e_{j,fk}^{\text{ex}}]_f$ modulated in time by $p_{j,kn}^{\text{ex}}$, i.e., $v_{j,fn}^{\text{ex}} = \sum_{k=1}^{K_j^{\text{ex}}} p_{j,kn}^{\text{ex}} e_{j,fk}^{\text{ex}}$ [9]. In order to further constrain the spectral fine structure of the spectral patterns, they are represented as linear combinations of L_j^{ex} narrowband spectral patterns $[w_{j,fl}^{\text{ex}}]_f$ [14], i.e., $e_{j,fk}^{\text{ex}} = \sum_{l=1}^{L_j^{\text{ex}}} u_{j,lk}^{\text{ex}} w_{j,fl}^{\text{ex}}$, where $u_{j,lk}^{\text{ex}}$ are non-negative weights. The series of time activation coefficients $p_{j,kn}^{\text{ex}}$ are also represented as sums of M_j^{ex} time-localized patterns, i.e., $p_{j,kn}^{\text{ex}} = \sum_{m=1}^{M_j^{\text{ex}}} h_{j,mn}^{\text{ex}} g_{j,km}^{\text{ex}}$. Altogether we have:

$$v_{j,fn}^{\text{ex}} = \sum_{k=1}^{K_j^{\text{ex}}} \sum_{m=1}^{M_j^{\text{ex}}} h_{j,mn}^{\text{ex}} g_{j,km}^{\text{ex}} \sum_{l=1}^{L_j^{\text{ex}}} u_{j,lk}^{\text{ex}} w_{j,fl}^{\text{ex}}, \tag{12}$$

and, introducing matrices $\mathbf{H}_j^{\text{ex}} \triangleq [h_{j,mn}^{\text{ex}}]_{m,n}$, $\mathbf{G}_j^{\text{ex}} \triangleq [g_{j,km}^{\text{ex}}]_{k,m}$, $\mathbf{U}_j^{\text{ex}} \triangleq [u_{j,lk}^{\text{ex}}]_{l,k}$ and $\mathbf{W}_j^{\text{ex}} \triangleq [w_{j,fl}^{\text{ex}}]_{f,l}$, this equation can be rewritten in matrix form as

$$\mathbf{V}_j^{\text{ex}} = \mathbf{W}_j^{\text{ex}} \mathbf{U}_j^{\text{ex}} \mathbf{G}_j^{\text{ex}} \mathbf{H}_j^{\text{ex}}. \tag{13}$$

3.5.3 Filter power structure

The filter spectral power $[v_{j,fn}^{\text{ft}}]_f$ is represented using exactly the same structure as in (12).

3.5.4 Total power structure

Altogether the spectral power structure can be represented by the hierarchical nonnegative matrix decomposition (3) (see also Tab. 2). Each matrix in the decomposition (3) is subject to specific constraints presented below.

3.5.5 Example constraints

Each matrix $\theta_{j,k}$ ($k = 2, \ldots, 9$) in (3) can be fixed, adaptive or partially fixed (see Tab. 3). In the latter two cases, a probabilistic prior $p(\theta_{j,k}|\eta_{j,k})$, such as a time continuity prior [9] or a sparsity-inducing prior [4] can be set. We denote by $\eta_{j,k}$ the *hyperparameters* of the prior that can be fixed or adaptive as well.

To cover *discrete state-based models* such as GMM, HMM, and their scaled versions GSMM, S-HMM, every column $\mathbf{g}_{j,m}^{\mathrm{ex}} = [g_{j,km}^{\mathrm{ex}}]_k$ of matrix $\mathbf{G}_j^{\mathrm{ex}}$ (and similarly for matrix $\mathbf{G}_j^{\mathrm{ft}}$) may further be constrained to have either a single nonzero entry (for GSMM, S-HMM) or a single nonzero entry equal to 1 (for GMM, HMM). Let $q_{j,m}^{\mathrm{ex}} \in \{1, \ldots, K_j^{\mathrm{ex}}\}$ be the index of the corresponding nonzero entry and $\mathbf{q}_j^{\mathrm{ex}} = [q_{j,m}^{\mathrm{ex}}]_m$ the resulting *state sequence* [5]. The prior distribution of $\theta_{j,4} = \mathbf{G}_j^{\mathrm{ex}}$ with hyperparameters $\eta_{j,4} = \mathbf{\Lambda}_j^{\mathrm{ex}}$ is defined as

$$p(\theta_{j,4}|\eta_{j,4}) = p(\mathbf{q}_j^{\mathrm{ex}}|\mathbf{\Lambda}_j^{\mathrm{ex}}) = \prod_{m=2}^{M_j^{\mathrm{ex}}} \lambda_{j,q_{j,m-1}^{\mathrm{ex}} q_{j,m}^{\mathrm{ex}}}^{\mathrm{ex}}, \qquad (14)$$

where $\mathbf{\Lambda}_j^{\mathrm{ex}} = [\lambda_{j,kk'}^{\mathrm{ex}}]_{k,k'}$ ($\lambda_{j,kk'}^{\mathrm{ex}} = \mathbb{P}(q_{j,m}^{\mathrm{ex}} = k'|q_{j,m-1}^{\mathrm{ex}} = k)$) denotes the $K_j^{\mathrm{ex}} \times K_j^{\mathrm{ex}}$ state transition probability matrix with $\lambda_{j,kk'}^{\mathrm{ex}}$ being independent on k (i.e., $\lambda_{j,kk'}^{\mathrm{ex}} = \lambda_{j,k'}^{\mathrm{ex}}$) in the case of GMM or GSMM.

4 Estimation algorithm

In this section we describe in details the proposed algorithm for the estimation of the model parameters and subsequent source separation.

4.1 Model estimation criterion

To estimate the model parameters, we use the standard maximum *a posteriori* (MAP) where the log-likelihood $\log p(\mathbf{x}_{fn}|\theta)$ in every TF point is replaced by its empirical expectation $\widehat{\mathbb{E}}[\log p(\mathbf{x}_{fn}|\theta)]$ according to the empirical expectation operator $\widehat{\mathbb{E}}[\cdot]$ introduced in section 3.2.2 [10,18]. Mathematically rigorous derivation of this criterion is given in Appendix A. This criterion consists in maximizing the *modified log-posterior* $\widehat{\mathcal{L}}(\theta, \eta|\mathbf{X}) \triangleq \widehat{\mathbb{E}}[\log p(\theta, \eta|\mathbf{X})]$, where $\mathbf{X} = \{\mathbf{x}_{fn}\}_{f,n}$, over the model parameters θ and the hyperparameters $\eta = \{\eta_{j,k}\}_{j,k=1}^{J,9}$. This quantity can be rewritten, using (5) and (7), as:

$$\widehat{\mathcal{L}}(\theta, \eta|\mathbf{X}) \overset{\mathrm{c}}{=} \widehat{\mathcal{L}}(\mathbf{X}|\theta) + \log p(\theta|\eta) =$$
$$\sum_{f,n} \widehat{\mathbb{E}}[\log N_c(\mathbf{x}_{fn}|0, \mathbf{\Sigma}_{\mathbf{x},fn})] + \log p(\theta|\eta), \quad (15)$$

where $\mathbf{\Sigma}_{\mathbf{x},fn} \triangleq \sum_{j=1}^J v_{j,fn}\mathbf{R}_{j,fn}$, $\widehat{\mathcal{L}}(\mathbf{X}|\theta) \triangleq \widehat{\mathbb{E}}[\log p(\mathbf{X}|\theta)]$ is the *modified log-likelihood* and "$\overset{\mathrm{c}}{=}$" denotes equality up to a constant. Using (6), the resulting

---

[5]Note that we consider here the state sequence $\mathbf{q}_j^{\mathrm{ex}}$ as a parameter to be estimated, and not as a latent variable one integrates over, as it is usually done for GMM / HMM parameter estimation. This is indeed to achieve the goal of generality by making the E-step of the GEM algorithm independent of the specified constraints.

criterion can be expressed as [13], [18]:

$$\theta^*, \eta^* = \arg\min_{\theta,\eta} \sum_{f,n} \left[\text{tr}\left(\boldsymbol{\Sigma}_{\mathbf{x},fn}^{-1} \widehat{\mathbf{R}}_{\mathbf{x},fn} \right) + \log |\boldsymbol{\Sigma}_{\mathbf{x},fn}| \right]$$
$$- \sum_{j,k=1}^{J,9} \log p(\theta_{j,k}|\eta_{j,k}). \quad (16)$$

We see that this criterion does not rely any more on the linear mixture representation $\mathbf{X}$, but only on the resulting empirical mixture covariances $\{\widehat{\mathbf{R}}_{\mathbf{x},fn}\}_{f,n}$.

4.2 Model estimation via a GEM algorithm

Given the model parameters $\theta = \{\theta_{j,k}\}_{j,k=1}^{J,9}$ specified in table 2 and the hyperparameters $\eta = \{\eta_{j,k}\}_{j,k=1}^{J,9}$ together with user-defined constraints and initial values, we minimize the criterion (16) using a GEM algorithm [20] that consists in iterating the following expectation (E) and maximization (M) steps (see Fig. 4):

- *E-step:* Compute the conditional expectation of the so-called *natural (sufficient) statistics*, given the observations $\mathbf{X}$ and the current parameters θ, η.

- *M-step:* Given the expectation of the natural statistics, update the parameters θ, η so as to increase the conditional expectation of the modified log-posterior of the so-called *complete data*. This step is implemented via a loop over all $J \times 9$ parameter subsets $\theta_{j,k}$ specified in table 2. Each subset, depending whether it is adaptive (partially adaptive) or fixed, is updated (partially updated) or not in turn using suitable update rules [13,9,14].

4.2.1 Preliminaries

Additive noise and simulated annealing As explained in [13], where a similar GEM algorithm is used, the mixing parameters $\mathbf{A}_{fn}$ (see Eq. (10)) updated via this GEM algorithm can become stuck into a suboptimal value. To overcome this issue, we use a form of *simulated annealing* proposed in [13], which consisting of adding to (10) a noise term whose variance is decreased by a fixed amount at each iteration. Thus, we assume that there is a $J+1$-th source with full-rank time-invariant spatial covariance $\boldsymbol{\Sigma}_{\mathbf{b},fn} = \sigma_f^2 \mathbf{I}_I = \mathbf{R}_{J+1,fn}$ and trivial spectral power ($v_{J+1,fn} = 1$) that represents a controllable additive isotropic noise $\mathbf{b}_{fn} = \mathbf{y}_{J+1,fn}$. Introducing this noise component leads to considering the noise covariance $\boldsymbol{\Sigma}_{\mathbf{b},fn}$ as part of the model parameters θ and to adding it to the mixing equation (10):

$$\mathbf{x}_{fn} = \mathbf{A}_{fn}\mathbf{s}_{fn} + \mathbf{b}_{fn}. \quad (17)$$

Complete data log-posterior and natural statistics We chose $\mathbf{Z} = \{\mathbf{X}, \mathbf{S}\}$ as the complete data, where $\mathbf{S} = \{\mathbf{s}_{fn}\}_{f,n}$, and the modified log-posterior of the

complete data can be written as:

$$
\begin{aligned}
\widehat{\mathcal{L}}(\theta, \eta | \mathbf{X}, \mathbf{S}) &\overset{c}{=} \widehat{\mathcal{L}}(\mathbf{X}|\mathbf{S}; \theta) + \widehat{\mathcal{L}}(\mathbf{S}|\theta) + \log p(\theta|\eta) \\
&\overset{c}{=} -\sum_{f,n} \mathrm{tr}\left[\boldsymbol{\Sigma}_{\mathbf{b},fn}^{-1} \left(\mathbf{R}_{\mathbf{x},fn} - \mathbf{A}_{fn}\mathbf{R}_{\mathbf{xs},fn}^{H} \right.\right. \\
&\qquad \left.\left. -\mathbf{R}_{\mathbf{xs},fn}\mathbf{A}_{fn}^{H} - \mathbf{A}_{fn}\mathbf{R}_{\mathbf{s},fn}\mathbf{A}_{fn}^{H} \right)\right] - \sum_{f,n} \log |\boldsymbol{\Sigma}_{\mathbf{b},fn}| \\
&\qquad - \sum_{j} R_j \sum_{f,n} d_{IS}(\xi_{j,fn}|v_{j,fn}) + \sum_{j,k=1}^{J,9} \log p(\theta_{j,k}|\eta_{j,k}), \quad (18)
\end{aligned}
$$

where $d_{IS}(x|y) = \frac{x}{y} - \log \frac{x}{y} - 1$ is the Itakura-Saito (IS) divergence [9], $v_{j,fn}$ are the entries of matrix $\mathbf{V}_j$ specified by (3), and $\mathbf{R}_{\mathbf{x},fn}$, $\mathbf{R}_{\mathbf{xs},fn}$, $\mathbf{R}_{\mathbf{s},fn}$ and $\xi_{j,fn}$ are defined as:

$$
\mathbf{R}_{\mathbf{x},fn} \triangleq \widehat{\mathbf{R}}_{\mathbf{x},fn} = \widehat{\mathbb{E}}[\mathbf{x}_{fn}\mathbf{x}_{fn}^{H}], \quad \mathbf{R}_{\mathbf{xs},fn} \triangleq \widehat{\mathbb{E}}[\mathbf{x}_{fn}\mathbf{s}_{fn}^{H}], \quad (19)
$$

$$
\mathbf{R}_{\mathbf{s},fn} \triangleq \widehat{\mathbb{E}}[\mathbf{s}_{fn}\mathbf{s}_{fn}^{H}], \quad \xi_{j,fn} \triangleq \frac{1}{R_j} \sum_{r=1}^{R_j} \widehat{\mathbb{E}}[|s_{jr,fn}|^2]. \quad (20)
$$

It can be easily shown from (18) that the family of functions $\{\exp \widehat{\mathcal{L}}(\mathbf{X}, \mathbf{S}|\theta)\}_\theta$ forms an *exponential family* [20,7], and the set $\mathbb{T}(\mathbf{X}, \mathbf{S}) = \{\mathbf{R}_{\mathbf{x},fn}, \mathbf{R}_{\mathbf{xs},fn}, \mathbf{R}_{\mathbf{s},fn}\}_{f,n}$ is a *natural (sufficient) statistics* [7] for this family. Given this result, we derive a GEM algorithm that is summarized below.

4.2.2 Conditional expectation of the natural statistics (E-step)

The conditional expectations of the natural statistics $\mathbb{T}(\mathbf{X}, \mathbf{S})$ are computed as follows:

$$
\begin{aligned}
\widehat{\mathbf{R}}_{\mathbf{xs},fn} &= \widehat{\mathbf{R}}_{\mathbf{x},fn}\boldsymbol{\Omega}_{\mathbf{s},fn}^{H}, &(21) \\
\widehat{\mathbf{R}}_{\mathbf{s},fn} &= \boldsymbol{\Omega}_{\mathbf{s},fn}\widehat{\mathbf{R}}_{\mathbf{x},fn}\boldsymbol{\Omega}_{\mathbf{s},fn}^{H} + (\mathbf{I}_R - \boldsymbol{\Omega}_{\mathbf{s},fn}\mathbf{A}_f)\boldsymbol{\Sigma}_{\mathbf{s},fn}, &(22)
\end{aligned}
$$

where

$$
\begin{aligned}
\boldsymbol{\Omega}_{\mathbf{s},fn} &= \boldsymbol{\Sigma}_{\mathbf{s},fn}\mathbf{A}_{fn}^{H}\boldsymbol{\Sigma}_{\mathbf{x},fn}^{-1}, &(23) \\
\boldsymbol{\Sigma}_{\mathbf{x},fn} &= \mathbf{A}_{fn}\boldsymbol{\Sigma}_{\mathbf{s},fn}\mathbf{A}_{fn}^{H} + \boldsymbol{\Sigma}_{\mathbf{b},fn}, &(24) \\
\boldsymbol{\Sigma}_{\mathbf{s},fn} &= \mathrm{diag}\left([\phi_{r,fn}]_{r=1}^{R} \right), &(25)
\end{aligned}
$$

and $\phi_{r,fn} = v_{j,fn}$ if and only if $r \in \mathcal{R}_j$, where $\mathcal{R}_j$ denotes the set of sub-source indices associated with source j in the vector $\mathbf{s}_{fn}$ (see section 3.4).

4.2.3 Update of the spatial covariances (M-step)

Unconstrained time-invariant mixing parameters We first consider the case where there are no probabilistic priors specified for the mixing parameters $\{\mathbf{A}_j\}_j$ and these parameters are time-invariant. Let $\mathcal{A} \subset \{1, \dots, R\}$ be a subset of indices of size $\mathcal{D} = \#(\mathcal{A})$. Below we denote by $\mathbf{A}_{fn}^{\mathcal{A}}$, $\widehat{\mathbf{R}}_{\mathbf{xs},fn}^{\mathcal{A}}$ and $\widehat{\mathbf{R}}_{\mathbf{s},fn}^{\mathcal{A}}$ the matrices of respective sizes $I \times \mathcal{D}$, $I \times \mathcal{D}$ and $\mathcal{D} \times \mathcal{D}$, that consist of the corresponding

entries of the matrices $\mathbf{A}_{fn}$, $\widehat{\mathbf{R}}_{\mathbf{xs},fn}$ and $\widehat{\mathbf{R}}_{\mathbf{s},fn}$, i.e., $\mathbf{A}_{fn}^{\mathcal{A}} = [\mathbf{A}_{fn}(i,r)]_{i=1,r\in\mathcal{A}}^{I}$, $\widehat{\mathbf{R}}_{\mathbf{xs},fn}^{\mathcal{A}} = [\widehat{\mathbf{R}}_{\mathbf{xs},fn}(i,r)]_{i=1,r\in\mathcal{A}}^{I}$, and $\widehat{\mathbf{R}}_{\mathbf{s},fn}^{\mathcal{A}} = [\widehat{\mathbf{R}}_{\mathbf{s},fn}(r,r')]_{r,r'\in\mathcal{A}}$. We also denote by $\overline{\mathcal{A}} = \{1,\ldots,R\}\backslash\mathcal{A}$ the complementary set. Let $\mathcal{C} \subset \{1,\ldots,R\}$ (resp. $\mathcal{I} \subset \{1,\ldots,R\}$) be the indices of convolutively (resp. instantaneously) mixed sources with adaptive mixing parameters. With these conventions the mixing parameters are updated as follows [6]:

$$\mathbf{A}_{fn}^{\mathcal{C}} = \left[\sum_{\tilde{n}} \left\{\widehat{\mathbf{R}}_{\mathbf{xs},f\tilde{n}}^{\mathcal{C}} - \mathbf{A}_{f\tilde{n}}^{\overline{\mathcal{C}}}\widehat{\mathbf{R}}_{\mathbf{s},f\tilde{n}}^{\overline{\mathcal{C}}}\right\}\right] \left[\sum_{\tilde{n}} \widehat{\mathbf{R}}_{\mathbf{s},f\tilde{n}}^{\mathcal{C}}\right]^{-1}, \qquad (26)$$

$$\mathbf{A}_{fn}^{\mathcal{I}} = \Re\left[\sum_{\tilde{f},\tilde{n}} \left\{\widehat{\mathbf{R}}_{\mathbf{xs},\tilde{f}\tilde{n}}^{\mathcal{I}} - \mathbf{A}_{\tilde{f}\tilde{n}}^{\overline{\mathcal{I}}}\widehat{\mathbf{R}}_{\mathbf{s},\tilde{f}\tilde{n}}^{\overline{\mathcal{I}}}\right\}\right] \Re\left\{\sum_{\tilde{f},\tilde{n}} \widehat{\mathbf{R}}_{\mathbf{s},\tilde{f}\tilde{n}}^{\mathcal{I}}\right\}^{-1}. \qquad (27)$$

Other constraints   Estimating time-varying mixing parameters without any priors does not make much sense in practice due to highly unconstrained nature of such the estimation. If the mixing parameters are given some Gaussian priors, closed-form updated similar to (26), (27) can be still derived, since the modified log-posterior (18) will be a quadratic form with respect to the mixing parameters. In case of nongaussian priors some Newton-like updates [21] can be derived.

4.2.4   Update of the spectral power parameters (M-step)

Unconstrained nonnegative matrices   Let $\mathbf{C}_j = \theta_{j,k}$ ($k = 2,\ldots,9$) an adaptive or partially adaptive nonnegative matrix (see Tab 2) with a uniform prior $p(\theta_{j,k}|\eta_{j,k}) = 1$. Whatever the matrix $\mathbf{C}_j$, it can be shown that the decomposition (3) can be rewritten as $\mathbf{V}_j = (\mathbf{B}_j\mathbf{C}_j\mathbf{D}_j) \odot \mathbf{E}_j$, where $\mathbf{B}_j$, $\mathbf{D}_j$ and $\mathbf{E}_j$ are some nonnegative matrices that are assumed to be fixed while $\mathbf{C}_j$ is updated. For example, if $\mathbf{C}_j = \mathbf{H}_j^{\text{ft}}$ in (3), one can choose $\mathbf{B}_j = \mathbf{W}_j^{\text{ft}}\mathbf{U}_j^{\text{ft}}\mathbf{G}_j^{\text{ft}}$, $\mathbf{D}_j = \mathbf{I}_N$ and $\mathbf{E}_j = \mathbf{W}_j^{\text{ex}}\mathbf{U}_j^{\text{ex}}\mathbf{G}_j^{\text{ex}}\mathbf{H}_j^{\text{ex}}$. With these notations it can be shown that the conditional expectation of the log-posterior (18) of the complete data is non-decreasing when the corresponding update for $\mathbf{C}_j$ does not increase the following cost function:

$$\mathcal{D}_{IS}(\mathbf{C}_j) = \sum_{f,n} d_{IS}([\widehat{\boldsymbol{\Xi}}_j]_{f,n}|[\mathbf{V}_j]_{f,n}), \qquad (28)$$

where $\mathbf{V}_j = (\mathbf{B}_j\mathbf{C}_j\mathbf{D}_j)\odot\mathbf{E}_j$ and $\widehat{\boldsymbol{\Xi}}_j = [\hat{\xi}_{j,fn}]_{f,n}$ with $\hat{\xi}_{j,fn}$ computed as follows:

$$\hat{\xi}_{j,fn} = \frac{1}{R_j} \sum_{r\in\mathcal{R}_j} \widehat{\mathbf{R}}_{\mathbf{s},fn}(r,r), \qquad (29)$$

where $\widehat{\mathbf{R}}_{\mathbf{s},fn}$ is computed in (22) and $\mathcal{R}_j$ is defined at the end of section 4.2.2. Applying some standard derivations (see, e.g., [9]), one can obtain the following

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[6] We see that the mixing parameters for different sources are updated jointly by Eqs. (26), (27), while we have claimed in the beginning of section 4 that they will be updated in an alternated manner. However, since we can here update parameters jointly without loss of flexibility, we do so, since joint optimization, as compared to the alternated one, leads in general to a faster convergence.

nonnegative multiplicative update (MU) rule [7]

$$\mathbf{C}_j = \mathbf{C}_j \odot \frac{\mathbf{B}_j^T [\widehat{\boldsymbol{\Xi}}_j \odot \mathbf{E}_j \odot \{(\mathbf{B}_j \mathbf{C}_j \mathbf{D}_j) \odot \mathbf{E}_j\}^{\cdot -2}] \mathbf{D}_j^T}{\mathbf{B}_j^T [\mathbf{E}_j \odot \{(\mathbf{B}_j \mathbf{C}_j \mathbf{D}_j) \odot \mathbf{E}_j\}^{\cdot -1}] \mathbf{D}_j^T} \tag{30}$$

that guarantees non-increase of the cost function (28), and thus non-decrease of the conditional expectation of the modified log-posterior (18) of the complete data. These update rules, as applied to multichannel audio, are in fact a generalization of the GEM-MU algorithm proposed in [40], that has been shown to converge much more quickly than the GEM algorithm in [13].

Discrete state-based constraints Let us now assume that $\theta_{j,4} = \mathbf{G}_j^{\text{ex}}$ is subject to a discrete state-based constraint (similarly for $\theta_{j,8} = \mathbf{G}_j^{\text{ft}}$). The updates are performed as follows:

1. Set $\tilde{\mathbf{G}}_j^{\text{ex}} = \mathbf{G}_j^{\text{ex}}$, and fill each entry of each column of $\tilde{\mathbf{G}}_j^{\text{ex}}$ with the nonzero entry of the respective column of $\mathbf{G}_j^{\text{ex}}$.

2. If $\mathbf{G}_j^{\text{ex}}$ is adaptive, do for every $k = 1, \ldots, K_j^{\text{ex}}$:

   - Set $\mathbf{C}_j = \tilde{\mathbf{G}}_j^{\text{ex}}$, and set all the elements of $\mathbf{C}_j$ to zero, except the k-th row.
   - Update $\mathbf{C}_j$ using several iterations of (30) [8].
   - Set the k-th row of $\tilde{\mathbf{G}}_j^{\text{ex}}$ equal to that of $\mathbf{C}_j$.

3. For every $k = 1, \ldots, K_j^{\text{ex}}$ and $m = 1, \ldots, M_j^{\text{ex}}$ set $\mathbf{C}_j = \tilde{\mathbf{G}}_j^{\text{ex}}$, set all the elements of $\mathbf{C}_j$ to zero, except the (k, m)-th one, and compute the IS divergence $\mathcal{D}_{IS}(k, m)$ between $\mathbf{V}_j = (\mathbf{B}_j \mathbf{C}_j \mathbf{D}_j) \odot \mathbf{E}_j$ and $\widehat{\boldsymbol{\Xi}}_j$, as in (28).

4. Update the state sequence $\mathbf{q}_j^{\text{ex}}$ using the Viterbi algorithm [41] to minimize the following criterion:

$$\mathbf{q}_j^{\text{ex}} = \arg \min_{\mathbf{q}_j^{\text{ex}}} \sum_{m=2}^{M_j^{\text{ex}}} \mathcal{D}_{IS}(q_{j,m}^{\text{ex}}, m) - \log p(\mathbf{q}_j^{\text{ex}} | \boldsymbol{\Lambda}_j^{\text{ex}}),$$

   where $p(\mathbf{q}_j^{\text{ex}} | \boldsymbol{\Lambda}_j^{\text{ex}})$ is computed as in (14).

5. Set $\mathbf{G}_j^{\text{ex}} = \tilde{\mathbf{G}}_j^{\text{ex}}$ and set to zero all the entries of $\mathbf{G}_j^{\text{ex}}$, except those corresponding to $\mathbf{q}_j^{\text{ex}}$.

6. If $\boldsymbol{\Lambda}_j^{\text{ex}}$ is adaptive, update the transition probabilities as

$$\lambda_{j,kk'}^{\text{ex}} = \frac{1}{(M_j^{\text{ex}} - 1) K_j^{\text{ex}}} \sum_{m=2}^{M_j^{\text{ex}}} \mathbf{1}(q_{j,m-1}^{\text{ex}} = k, q_{j,m}^{\text{ex}} = k')$$

---

[7] In the case of partially adaptive matrix $\mathbf{C}_j$, only the adaptive matrix entries are updated with rule (30).

[8] Several iterations of update rule (30) are needed because all entries of $\tilde{\mathbf{G}}_j^{\text{ex}}$ are initialized in step 1 from a particular sequence of gains carried by $\mathbf{G}_j^{\text{ex}}$ and optimized for the current state sequence $\mathbf{q}_j^{\text{ex}}$. Performing only one update of (30) would unfavor state sequence evaluation. However, to avoid all these issues, in our implementation we just keep matrix $\tilde{\mathbf{G}}_j^{\text{ex}}$ in memory, skip step 1, and do only one iteration of (30).

in case of HMM or S-HMM or as

$$\lambda_{j,kk'}^{\mathrm{ex}} = \frac{1}{M_j^{\mathrm{ex}} - 1} \sum\nolimits_{\substack{m=2}}^{M_j^{\mathrm{ex}}} \mathbf{1}_{(q_{j,m}^{\mathrm{ex}}=k')}$$

in case of GMM or GSMM.

4.2.5 Other constraints

We here discuss the updates that are not yet included in our current baseline implementation (see Sec. 2.5).

An EM algorithm update rules for time pattern weights $\mathbf{G}_j^{\mathrm{ex}}$ or $\mathbf{G}_j^{\mathrm{ft}}$ with time continuity priors, such as inverse-Gamma or Gamma Markov chain priors, can be found in [9]. However, one cannot use these rules within our framework, since we use a different, reduced, complete data set, as compared to the one used in [9]. Nevertheless, one can always use some Newton-like updates [21] for these priors.

If a matrix $\theta_{j,k}$ $(k = 2, \ldots, 9)$ is constrained with a sparsity-inducing prior [4], such as a Laplacian prior (corresponding to an l_1 norm penalty), it can be updated using the multiplicative updates described in [42, 43]. However, in such a case the renormalization described in the subsection below could not be applied, since it would change the value of the optimized criterion (16). At the same time, without any renormalization, the sparsity-inducing prior would loose its influence. To avoid that, all the other parameter subsets $\theta_{j,l}$ $(l \neq k)$ should be constrained, e.g., to have a unitary (say l_1) norm, which can be handled using the gradient descent updates from [42] or the modified multiplicative updates from [43].

4.2.6 Renormalization

At the end of each GEM iteration, in order to avoid numerical (under/over-flow) problems, a renormalization of some parameters is done if needed, i.e., if these parameters are not already constrained by some priors that are not scale-invariant. This procedure is similar to the one described in [13], and it does not change the value of the optimized criterion (16). For example, the columns of matrix $\mathbf{U}_j^{\mathrm{ex}}$ can be divided by their energies, and the rows of $\mathbf{G}_j^{\mathrm{ex}}$ scaled accordingly (see (3)). Similar renormalization is applied in turn to each patameter subsets pairs $\theta_{j,k}, \theta_{j,k+1}$ $(k = 1, \ldots, 8)$, and at the end of this operation the total energy is relegated into $\theta_{j,9}$.

4.3 Source estimation

Given the estimated model parameters θ, the sources can be estimated in the minimum mean square error (MMSE) sense via the Wiener filtering:

$$\hat{\mathbf{y}}_{j,fn} = v_{j,fn} \mathbf{R}_{j,fn} \mathbf{\Sigma}_{\mathbf{x},fn}^{-1} \mathbf{x}_{fn}, \tag{31}$$

where $\mathbf{\Sigma}_{\mathbf{x},fn} = \sum_{j=1}^{J} v_{j,fn} \mathbf{R}_{j,fn}$. The counterpart of this equation for quadratic TF representations is given in Appendix A.

5 Experimental Illustrations

The goals of this experimental part are to illustrate on some examples how to specify the prior information in the framework, given a particular source separation problem, and to demonstrate that we can implement the existing and new methods within the framework. For that we first give an example of application of the framework to a music recording in a *non-blind* setting, i.e., when different sources are given different models according to the prior information. Second, we consider a few blind framework instances, corresponding to existing and new methods, and apply them for separation of underdetermined speech and music mixtures.

5.1 Non-blind separation of one music recording

5.1.1 Data

As an example stereo music recording to separate we took the 23-second snip of the song "Que pena tanto faz" by Tamy from the test dataset of the SiSEC 2008 [28] "Professionally produced music recordings" task. We know about this recording that there are two sources, a female singing voice and a guitar, that the voice is instantaneously mixed (panned) in the middle [9] and the guitar is possibly a non-point convolutive source.

5.1.2 Constraint specification and parameter initialization

To account for this information within our framework, we have chosen the following constraints. The singing voice mixing parameters $\mathbf{A}_1$ form a fixed tensor of size $2 \times 1 \times F \times N$ with all entries equal to 1. The guitar mixing parameters $\mathbf{A}_2$ form an adaptive tensor of size $2 \times 2 \times F \times N$ subject to time-invariance constraint. The spectral powers $\mathbf{V}_j$ $(j = 1, 2)$ are constrained to $\mathbf{V}_j = \mathbf{W}_j^{\text{ex}} \mathbf{U}_j^{\text{ex}} \mathbf{G}_j^{\text{ex}} \mathbf{H}_j^{\text{ex}}$ [3] with $\mathbf{W}_j^{\text{ex}}$ and $\mathbf{H}_j^{\text{ex}}$ being fixed, and $\mathbf{U}_j^{\text{ex}}$ and $\mathbf{G}_j^{\text{ex}}$ being adaptive. The narrowband spectral patterns $\mathbf{W}_j^{\text{ex}}$ include $6 \times L$ harmonic patterns modeling the harmonic part of L pitches and 9 smooth patterns (see Fig. 3 (E) and [14]). The L pitches are chosen to cover the range of 77 - 1500 Hz (39 - 89 on the MIDI scale), which is enough for both the guitar and this particular singing. The time-localized patterns $\mathbf{H}_1^{\text{ex}}$ and $\mathbf{H}_2^{\text{ex}}$ are different. The singing voice time-localized patterns $\mathbf{H}_1^{\text{ex}}$ include half-Gaussians truncated at the left, i.e., only the right half is kept. The guitar time-localized patterns $\mathbf{H}_2^{\text{ex}}$ include decreasing exponentials to model the decay part of the notes and discrete Dirac functions to model note attacks (see Fig. 3 (H)). All adaptive parameters are initialized with random values. Finally, we used the ERB quadratic representation described in [18] as signal representation.

5.1.3 Results

After 500 iterations of the proposed GEM algorithm the separation results, measured in terms of the source to distortion ratio (SDR) [44], were 7.2 and 8.9 dB for voice and guitar, respectively. We have also separated the same mixture using all the blind settings described in the following section. The best results

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[9] This information can be for example obtained by subtracting the left channel from the right one and checking that the voice is cancelled.

of 5.5 and 7.1 dB SDR were obtained by the unconstrained NMF spectral power model with the instantaneous rank-1 mixing, i.e., by the multichannel NMF for instantaneous mixtures [13].

5.1.4 Discussion

We see that our informed setting outperforms any blind setting by at least 1.7 dB SDR. This improvement is essentially due to the combination of rank-1 instantaneous and full-rank convolutive mixing models and the information about the position of one source. Moreover, while it is common in professionally produced music recordings that some sources are mixed instantaneously (panned) and others convolutively (e.g., live-recorded tracks or some artificial reverberation is added), in our best knowledge such hybrid models were not yet proposed for audio source separation, and it now becomes possible to implement them within our framework.

5.2 Blind separation of underdetermined speech and music mixtures

5.2.1 Data

Here we evaluate several settings of our framework on the development dataset of the SiSEC 2010 [27] "Underdetermined-speech and music mixtures" task. This dataset include 10-seconds length instantaneous, convolutive and live-recorded stereo mixtures of three or four music and speech sources (see [27] for more details).

5.2.2 Constraint specification and parameter initialization

We consider eight blind settings of the framework that are specified by the following constraints. For all settings and for all sources $\mathbf{A}_j$ forms an adaptive tensor of size $2 \times R_j \times F \times N$ subject to time-invariance constraint and subject to frequency invariance constraint for instantaneous mixtures only. The spectral power of each source is structured as $\mathbf{V}_j = \mathbf{E}_j^{\mathrm{ex}} \mathbf{P}_j^{\mathrm{ex}}$. The eight settings are generated by all possible combinations of the following possibilities (see also Table 4):

- *Rank:* The rank R_j is either 1 or 2 (full-rank).

- *Spectral structure:* The characteristic spectral patterns $\mathbf{E}_j^{\mathrm{ex}}$ are either *unconstrained*, i.e., $\mathbf{E}_j^{\mathrm{ex}} = \mathbf{W}_j^{\mathrm{ex}}$ with adaptive $\mathbf{W}_j^{\mathrm{ex}}$, or *constrained*, i.e., $\mathbf{E}_j^{\mathrm{ex}} = \mathbf{W}_j^{\mathrm{ex}} \mathbf{U}_j^{\mathrm{ex}}$ with fixed $\mathbf{W}_j^{\mathrm{ex}}$ being composed of harmonic and noise-like and smooth narrowband spectral patterns (see Fig. 3 (E) and [14]), and adaptive $\mathbf{U}_j^{\mathrm{ex}}$ (see Fig. 3 (F)) that is very sparse so as to eliminate invalid combinations of narrowband spectral patterns (e.g., patterns corresponding to different pitches should not be combined together).

- *Temporal structure:* The time activation coefficients $\mathbf{P}_j^{\mathrm{ex}}$ are either *unconstrained*, i.e., $\mathbf{E}_j^{\mathrm{ex}} = \mathbf{G}_j^{\mathrm{ex}}$ with adaptive $\mathbf{G}_j^{\mathrm{ex}}$, or *constrained*, i.e., $\mathbf{E}_j^{\mathrm{ex}} = \mathbf{G}_j^{\mathrm{ex}} \mathbf{H}_j^{\mathrm{ex}}$ with fixed $\mathbf{H}_j^{\mathrm{ex}}$ being composed of decreasing exponentials, as those on Fig. 3 (H), and adaptive $\mathbf{G}_j^{\mathrm{ex}}$.

The two settings with $R_j = 1$ and 2, and unconstrained $\mathbf{E}_j^{\mathrm{ex}}$ and $\mathbf{P}_j^{\mathrm{ex}}$ correspond to the state-of-the-art methods [13] and [17], respectively (see Sec. 2.4), while the remaining six settings are new.

In line with [13], parameter estimation via GEM is sensitive to initialization for all the settings we consider. To provide our GEM algorithm with a "good initialization" we used for the instantaneous mixtures the DEMIX mixing matrix estimation algorithm [45] to initialize mixing parameters $\mathbf{A}_j$, followed by l_0 norm minimization (see e.g., [1]) and Kullback-Leibler (KL) divergence minimization (see [13]) to initialize the source power spectra $\mathbf{V}_j$. For synthetic convolutive and live recorded mixtures we first estimated the time differences of arrival (TDOAs) using the MVDRW estimation algorithm proposed in [46], that is based on a variance distortionless response (MVDR) beamformer. The estimated TDOAs were then used to initialize anechoic mixing parameters $\mathbf{A}_j$, followed by binary masking and KL divergence minimization (see [13]) to initialize the source power spectra $\mathbf{V}_j$. As signal representation we used the STFT.

5.2.3 Results

Source separation results in terms of average SDR after 200 iterations of the proposed GEM algorithm are summarized in table 4 together with results of the *baseline* used for initialization.

Mixing				instantaneous		synthetic convolutive				live recorded			
Sources				speech	music	speech		music		speech		music	
Microphone spacing				-	-	5 cm	1 m	5 cm	1 m	5 cm	1 m	5 cm	1 m
Number of 10 second-length mixtures				6	4	10	10	4	4	10	10	4	4
baseline (l_0 minimization [47] or binary masking)				8.6	12.4	1.0	1.4	-0.9	-0.7	1.1	1.4	2.5	0.3
Method	rank R_j	spectral struct.	temporal struct.										
[13]	1	unconstrained	unconstrained	8.8	17.2	1.6	2.1	-1.1	-1.2	2.2	2.5	3.2	0.4
[17]	2	unconstrained	unconstrained	8.9	17.0	1.8	2.7	-0.5	-0.2	2.0	3.0	3.5	0.8
new	1	constrained	unconstrained	**10.5**	13.6	1.9	2.5	-0.5	-0.5	2.2	2.8	3.0	0.5
new	2	constrained	unconstrained	10.4	13.0	**2.1**	3.1	-0.7	-0.4	2.3	3.2	3.2	0.8
new	1	unconstrained	constrained	8.9	**18.6**	1.5	2.2	-0.8	-0.5	2.4	2.6	3.4	0.9
new	2	unconstrained	constrained	8.7	15.4	1.8	2.6	-0.4	0.0	2.1	2.9	**4.5**	**1.8**
new	1	constrained	constrained	**10.5**	15.7	**2.1**	2.9	-1.2	**0.3**	**2.5**	3.9	3.2	0.4
new	2	constrained	constrained	10.2	13.8	**2.1**	**4.5**	**0.0**	-0.3	2.3	**5.0**	3.7	1.0

Table 4: Average SDRs on subsets of SiSEC 2010 "Underdetermined speech and music mixtures" task development dataset.

5.2.4 Discussion

As expected, in most cases rank-1 spatial covariances perform the best for instantaneous mixtures and full-rank spatial covariances perform the best for synthetic convolutive and live recorded mixtures. Moreover, in all the cases there is at least one of the six new methods that outperforms the state-of-the-art methods [13] and [17]. One can note that for music sources constraining the spectral structure does not improve the separation performance [10], however, constraining the temporal structure does improve it. For speech sources constraining both the spectral and the temporal structures improves the separation performance in most cases. This is probably because the unconstrained NMF model is a poor model for speech, due to vibrato effect, thus it cannot be estimated in a robust way from these quite short 10-second length mixtures. Introducing spectral and temporal constraints makes model estimation more robust.

6 Conclusion

We have introduced a general flexible audio source separation framework that generalizes several existing source separation methods, brings them into a common framework, and allows to imagine and implement new efficient methods, given the prior information about a particular source separation problem. Besides the framework itself, we proposed a new temporal structure for NMF-like decompositions and an original mixing model formulation combining rank-1 and full-rank spatial mixing models in a homogeneous way. Finally, we provided a proper probabilistic formulation of local Gaussian modeling for quadratic time-frequency representations.

In the experimental part we have illustrated how to specify the prior information about a particular source separation problem within the framework, and we have shown that the framework allows implementing existing and new efficient source separation methods. We have also demonstrated that in some situations our new propositions can improve the source separation performance, as compared to the state-of-the-art. As such combining instantaneous rank-1 and and convolutive full-rank can be useful for separation of professionally produced music recordings, and the newly proposed temporal structure for NMF-like decompositions brings some improvement for blind separation of underdetermined mixtures of speech and music sources.

As for further research, the following extensions could be introduced to the framework. In a similar fashion as for spectral power, a flexible structure can be specified for the mixing parameters. E.g., the time-varying mixing parameters could be represented in terms of time-localized and locally time-invariant mixing parameter patterns, thus allowing the modeling of moving sources. Another interesting extension would be to introduce possible coupling between parameter subsets, thus allowing, e.g., the representation of the characteristic spectral patterns of different sources as linear combinations of eigenvoices [48] or eigeninstruments [49]. In fact, some parameter subsets corresponding to different sources can share common properties, and introducing such a coupling would make the estimation of these parameters more robust.

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[10] The results for synthetic convolutive mixtures of music sources are not very informative because of the poor overall performance.

A Probabilistic formulation of the local Gaussian model for quadratic representations

Here we give a proper probabilistic formulation of the local Gaussian model (7) for quadratic representations, explaining the exact meaning of the empirical covariance (6) and a justification of the criterion (16).

A.1 Input representation

Following [10, 18], we assume that the considered quadratic TF representation is computed by local averaging of a linear TF representation such as a STFT or an ERB filterbank. We assume that the indexing of the considered linear TF complex-valued representation, hereafter noted as $m = 1, \ldots, M$, can be in general different from the indexing f, n of the quadratic representation (6). Such a formulation allows considering linear and quadratic representations with different TF resolutions, but also using linear TF representations that do not allow any uniform TF indexing, e.g., an ERB representation with different sampling frequencies in different frequency bands or a signal-adapted multiple-window STFT [50]. The mixing equation (4) now writes as

$$\mathbf{x}_m = \sum\nolimits_{j=1}^{J} \mathbf{y}_{j,m}, \tag{32}$$

and we re-define the empirical covariance (6) as

$$\widehat{\mathbf{R}}_{\mathbf{x},fn} = \sum\nolimits_{m} (\omega_{fn,m}^{\mathrm{ana}})^2 \mathbf{x}_m \mathbf{x}_m^H, \tag{33}$$

where $\omega_{fn,m}^{\mathrm{ana}} \geq 0$, satisfying $\sum_{f,n} (\omega_{fn,m}^{\mathrm{ana}})^2 = 1$, are the coefficients of a local bi-dimensional *analysis* window specifying a neighbourhood of the TF point (f, n) [10, 18].

A.2 Local Gaussian model

In this setting the local Gaussian model (7) is re-defined as follows. Each vector $\mathbf{y}_{j,m}$ is assumed to be distributed as

$$\mathbf{y}_{j,m} \sim \mathcal{N}_c \left(\bar{0}, v_{j,fn} \mathbf{R}_{j,fn} \right) \tag{34}$$

with probability $(\omega_{fn,m}^{\mathrm{ana}})^2$. In other words, $\mathbf{y}_{j,m}$ is a realization of a GMM. Moreover, the vectors $\{\mathbf{y}_{j,m}\}_j$ are assumed to be independent only conditionally on the same GMM state. More precisely, the joint probability density function of $\{\mathbf{y}_{j,m}\}_j$ is defined as

$$p(\mathbf{y}_{1,m}, \ldots, \mathbf{y}_{J,m}) \triangleq \sum\nolimits_{fn} (\omega_{fn,m}^{\mathrm{ana}})^2 \prod\nolimits_j N_c \left(\mathbf{y}_{j,m}; \bar{0}, v_{j,fn} \mathbf{R}_{j,fn} \right). \tag{35}$$

A.3 Model estimation criterion

Under the above-presented assumptions (see (32) and (35)), the log-posterior $\log p(\theta, \eta | \mathbf{X})$, maximized by the MAP criterion, writes

$$\log p(\theta, \eta | \mathbf{X}) \stackrel{c}{=} \log p(\mathbf{X} | \theta) + \log p(\theta | \eta) =$$
$$\sum_{f,n} \log \sum_m (\omega_{fn,m}^{\mathrm{ana}})^2 N_c(\mathbf{x}_m; \bar{0}, \mathbf{\Sigma}_{\mathbf{x},fn}) + \log p(\theta | \eta), \tag{36}$$

where $\boldsymbol{\Sigma}_{\mathbf{x},fn} = \sum_{j=1}^{J} v_{j,fn}\mathbf{R}_{j,fn}$. Log-posterior (36) is difficult to optimize, due to summations in log-domain. Thus, following the EM methodology [20], we replace $\log p(\theta, \eta | \mathbf{X})$ by its lower bound

$$\sum_{f,n} \sum_{m} (\omega_{fn,m}^{\mathrm{ana}})^2 \log N_c(\mathbf{x}_m; \bar{0}, \boldsymbol{\Sigma}_{\mathbf{x},fn}) + \log p(\theta | \eta), \qquad (37)$$

using Jensen's inequality [20], and we get the criterion (16) with empirical covariances $\widehat{\mathbf{R}}_{\mathbf{x},fn}$ computed as in (33). Thus, the criterion (16) maximizes a lower bound of the log-posterior (36).

Note, that with this formulation we could obtain exactly the same updates as those presented in section 4.2 by deriving a GEM algorithm for the MAP criterion (36). This is because the computing of the lower bound (37) is based on the EM methodology. However, we prefer to keep the criterion (16), since it makes the formulation more compact and links it to quadratic representations and to the existing works [10, 18].

A.4   Source estimation

The sources can be estimated as follows [10, 18]:

$$\hat{\mathbf{y}}_{j,m} = \sum_{f,n} \omega_{fn,m}^{\mathrm{syn}} \omega_{fn,m}^{\mathrm{ana}} v_{j,fn} \mathbf{R}_{j,fn} \boldsymbol{\Sigma}_{\mathbf{x},fn}^{-1} \mathbf{x}_m, \qquad (38)$$

where $\omega_{fn,m}^{\mathrm{syn}} \geq 0$ is a so-called *synthesis* window satisfying $\sum_{f,n} \omega_{fn,m}^{\mathrm{syn}} \omega_{fn,m}^{\mathrm{ana}} = 1$. This estimator becomes the MMSE estimator when $\omega_{fn,m}^{\mathrm{syn}} = \omega_{fn,m}^{\mathrm{ana}}$.

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