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A Refinement-Based Validation Method for Programmable Logic Controllers

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Abstract—Programmable logic controllers (PLCs) are widely used in computer-based industrial applications. Timers play a pivotal role in PLC real-time embedded system applications. The paper addresses the formal validation of PLC systems with timers in the theorem proving system Coq. The timer behavior is characterized formally. A refinement validation methodology is presented in terms of an abstract model and a concrete model. The refinement is calibrated by a mapping relation. The soundness of the methodology is shown in the proving system. An illustrative case study demonstrates the effectiveness of the approach.

I. INTRODUCTION

Programmable logic controllers (PLCs) are widely used for safety critical applications in various industrial fields. The correctness and reliability of PLC programs are of great importance. Two distinguished features of PLCs programs are the use of timers and the cyclic behavior. In our previous work [1], we formally modeled and proved the correctness of a PLC program with timer control. The behavior of the program is represented by a set of sequences of systems states. Timers are explicitly modeled by a set of axioms over sequences. The model is at the scan cycle level, which means that at the beginning of each scan cycle, a system state is sampled to form the sequence. When compared to the rung level model (i.e., a system state is sampled before the execution of each rung to form a sequence), the scan-cycle level model is an abstract model. In [1], five assumptions and constraints are proposed to ensure the correctness of the abstract model (scan-cycle level model), but the correctness is not formally proved. In this paper, we continue our work: the correctness of the abstract model is formally proved in the theorem proving system Coq based on translation validation.

Translation validation [2] is an approach to the verification of translators. It is deployed to check whether the generated code is a correct implementation of the source code. In this paper we adopt this approach to prove the correctness of the abstract model according to a concrete model. The key to prove the correctness is to find a refinement relation between the two models. In order to establish the refinement relation, as described in [2], we need a common semantic framework for the representations of models is needed and a formal definition of refinement relation. Because of the real-time nature and time related properties of PLC programs, we

adopt an extension of *synchronous transition system* that can express time. We use timed execution sequences [3] as the common semantic base for the concrete and abstract models. The behavior of the concrete (or abstract) model is modeled by a set of timed execution sequences. The notion of refinement is then defined as a relation among sequences from the two sets. In this paper, the five assumptions and constraints are formally defined. And it is shown that they are enough to ensure the correctness of the abstract model.

The rest of the paper is organized as follows. In section I-A we summarize the related works. The formal definitions of the computational model and refinement relation is given in section II. The principles of modeling PLC programs with timers as well as the outline of how to prove the refinement relation are described in detail in section III. Finally section IV concludes the paper.

A. Related Works

There are already a lot of works dealing with the modeling and verification of PLC programs [4][5][6]. To the best of our knowledge, the work that is most close to ours is [7]. In [7], the authors employ translation validation for code-generators of PLCs. Compared to their work, our definition of refinement relation is different in two ways. First they do not have the sampling function. The introduction of sampling function helps to build a refinement relation between two models and ensure that the cyclic behavior of PLCs is modeled. Second they have a constraint that the initial states of the abstract model are mapped to a subset of the initial states of the concrete model.

II. THE COMPUTATION MODEL AND THE REFINEMENT RELATION

Let \mathcal{V} be a set of variables. A *state* s is a type-consistent valuation of variables in \mathcal{V} . Given a variable $v \in \mathcal{V}$ and a state s , $s[v]$ denotes the value of v at s . The set of all states over \mathcal{V} is denoted by Σ . We adopt the definition of synchronous transition systems in [2] with some modifications.

Definition 1: The following components define a *synchronous transition system* (STS) $S = (V, \Theta, \rho)$:

- $V \subseteq \mathcal{V}$: A finite set of system variables.
- Θ : An initial condition predicate over system states.

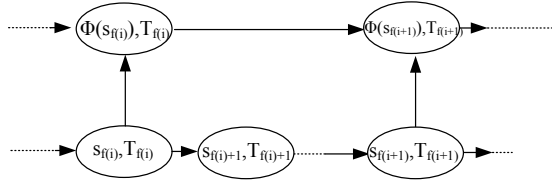


Fig. 1. The refinement relation between sequences

- ρ : A transition relation. ρ is a predicate over two states. Given two states s and s' , $\rho(s, s') = true$ iff s' is a possible successor state of s .

Before we introduce the additional time-related predicates, we define the timed state sequences.

Definition 2: A *timed state sequence* is a 2-tuple (σ, T) where σ an infinite state sequence and T is an infinite time sequence satisfying *monotonicity*, i.e., $\forall i \geq 0 T_{i+1} \geq T_i$, and *progress*, i.e., $\forall n \in \mathbb{N}, \exists i \geq 0, T_i > n$.

A *time related predicate* is a predicate over timed state sequences.

Definition 3: A *timed synchronous transition system (TSTS)* is a 2-tuple (S, P) where S is an STS and P is a set of time related predicates.

Given a TSTS $A = (S, P)$ where $S = (V, \Theta, \rho)$, a *timed execution sequence* of A is a timed state sequence (σ, T) satisfying :

- *Initiation* : $\Theta(\sigma_0)$ holds.
- *Consecution* : $\forall i \in \mathbb{N}, \rho(\sigma_i, \sigma_{i+1})$ holds.
- *Timing Relation* : $\forall p \in \mathcal{P}, p(\sigma)$ holds.

The set of all timed execution sequences is denoted by $\|A\|$.

A PLC program can be modeled by a TSTS. The non-time-related part of the program can be modeled by an STS, while the time-related part can be modeled by a set of time related predicates.

The refinement relation between two TSTSs is given below based on their timed execution sequences.

Definition 4: Given two TSTSs A and C , and two functions: a sampling function $f : \mathbb{N} \rightarrow \mathbb{N}$ satisfying $\forall i : \mathbb{N}, f(i) < f(i+1)$ and a state mapping function $\phi : \sum_C \rightarrow \sum_A$, C is a *refinement of A relative to f and ϕ* iff $\forall (\sigma_C, T_C) \in \|C\|, \exists (\sigma_A, T_A) \in \|A\|, \forall i \in \mathbb{N}, \phi(\sigma_C(f(i))) = \sigma_A(i) \wedge T_C(f(i)) = T_A(i)$.

Definition 5 is illustrated graphically in Fig. 1. Intuitively, given f, ϕ as described above and a timed execution sequence (σ, T) of the concrete model, $\{f(0), f(1), \dots\}$ are the sampling points along the concrete sequence. After sampling, we obtain a timed state sequence $(\sigma_{f(0)}\sigma_{f(1)} \dots, T_{f(0)}T_{f(1)} \dots)$. By applying ϕ to each state of the first part of the timed state sequence, we would obtain a timed execution sequence $(\phi(\sigma_{f(0)})\phi(\sigma_{f(1)}) \dots, T_{f(0)}T_{f(1)} \dots)$ of the abstract model.

Definition 5: For systems A and C , A refines B iff there exists two functions f and ϕ such that C is a refinement of A relative to f and ϕ .

The problem of checking the correctness of the abstract model turns to a problem of finding a proper sampling function and a proper state mapping function between the abstract

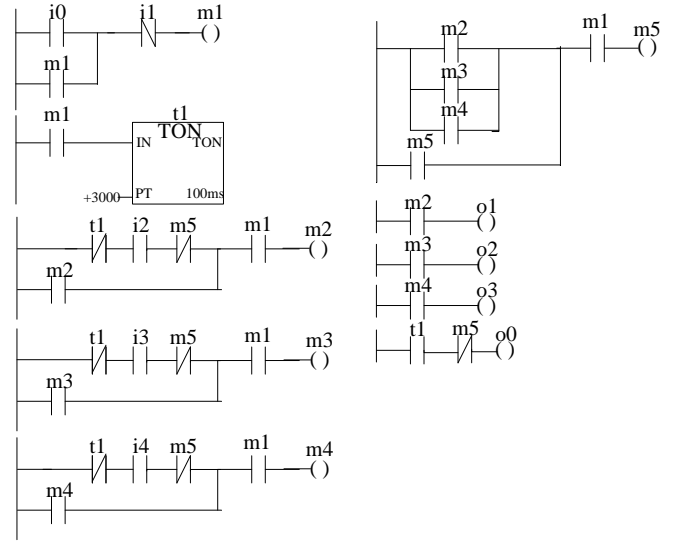


Fig. 2. The Example Ladder Diagram

model and the concrete model and prove they satisfy the property in definition 5.

III. MODELING AND VERIFICATION OF PLC PROGRAMS WITH TIMERS

In this section, we show how to build a rung level model of the example program and how to prove the soundness of the scan-cycle level model according to the rung level model. The example PLC program is shown in Fig. 2. For an introduction of PLC, timers and the scan-cycle level model of the example program, because of the limit of space, please refer to [1].

A. Modeling PLC Programs

1) *Variables:* All relays used in the program are boolean, so they are modeled by variables of type $nat \rightarrow Prop$. Since the model is at the rung level, there is a special variable pc (i.e., the program counter) of type $nat \rightarrow location$ that is used to indicate the current rung under execution. $location$ is inductively defined, but it can be understood as a set $\{e_1 \dots e_{11}\}$ among which e_{11} denotes the input/output action and for each $i \in \{1 \dots 10\}$, e_i represents the i -th rung in the example program.

```
Inductive location : Set :=
  e11|e1|e2|e3|e4|e5|e6|e7|e8|e9|e10.
Definition Var := nat -> Prop.
Variable pc : nat -> location.
```

Then, we have the following definitions of variables. Each variable corresponds to a relay of the example program.

```
Variables i0 i1 i2 i3 i4 m1 m2
  m3 m4 m5 o0 o1 o2 o3 t1: Var.
```

2) *Non-Time-Related Part:* As mentioned before, the non-time-related part can be modeled by an STS. An STS consists of two parts : the initiation condition and the consecution condition. The initiation condition $initial_states_c$ states the property that the first state of a sequence should satisfy. And the consecution condition $next_states_c$ expresses the property that two adjacent states should satisfy.

```
Definition initial_condition_c
  i0 i1 i2 i3 i4 m1 m2 m3 m4 m5 o0 o1 o2 o3 pc t1:=
```

```

 $\sim i_0 \wedge \sim i_1 \wedge \sim i_2 \wedge \sim i_3 \wedge \sim i_4 \wedge \sim m_1 \wedge \sim m_2 \wedge \sim m_3 \wedge \sim m_4 \wedge \sim m_5$ 
 $\wedge \sim o_0 \wedge \sim o_1 \wedge \sim o_2 \wedge \sim o_3 \wedge pc = e_{11} \wedge \sim t_1.$ 

```

Definition next_condition_c

```

(i0 i1 i2 i3 i4 m1 m2 m3 m4 m5 o0 o1 o2 o3:Prop) l
(t1 i0' i1' i2' i3' i4' m1' m2' m3' m4' m5' o0'
o1' o2' o3':Prop) l' t1':=

```

match l with

```

e1 => l' = e1 / \pres(V\{i0,i1,i2,i3,l})
| e1 => m1' = ((m1 \ \sim i0) \ \sim i1) \ \wedge l' = e2 / \pres(V\{m1,l})
| e2 => l' = e3 / \pres(V\{t1,l})
| e3 => m2' = ((m2 \ (\sim t1 \ \wedge \sim m5)) \ \wedge m1) \ \wedge l' = e4
\ \wedge \pres(V\{m2,l})
| e4 => m3' = ((m3 \ (\sim t1 \ \wedge \sim m5)) \ \wedge m1) \ \wedge l' = e5
\ \wedge \pres(V\{m3,l})
| e5 => m4' = ((m4 \ (\sim t1 \ \wedge \sim m5)) \ \wedge m1) \ \wedge l' = e6
\ \wedge \pres(V\{m4,l})
| e6 => m5' = ((m2 \ m3 \ m4 \ m5) \ \wedge m1) \ \wedge l' = e7
\ \wedge \pres(V\{m5,l})
| e7 => o1' = m2 \ \wedge l' = e8 / \pres(V\{o1,l})
| e8 => o2' = m3 \ \wedge l' = e9 / \pres(V\{o2,l})
| e9 => o3' = m4 \ \wedge l' = e10 / \pres(V\{o3,l})
| e10 => o0' = (t1 \ \sim m5) \ \wedge l' = e11 / \pres(V\{o0,l})

```

end.

Hypothesis initial_states_c : initial_condition_c

```

(i0 0)(i1 0)(i2 0)(i3 0)(i4 0)(m1 0)(m2 0)(m3 0)
(m4 0)(m5 0)(o0 0)(o1 0)(o2 0)(o3 0)(pc 0)(t1 0).

```

Hypothesis next_states_c : forall p,

```

next_condition_c (i0 p)(i1 p)(i2 p)(i3 p)(i4 p)
(m1 p)(m2 p)(m3 p)(m4 p)(m5 p)(o0 p)(o1 p)(o2 p)
(o3 p)(pc p)(t1 p)(i0 (S p))(i1 (S p))(i2 (S p))
(i3 (S p))(i4 (S p))(m1 (S p))(m2 (S p))(m3 (S p))
(m4 (S p))(m5 (S p))(o0 (S p))(o1 (S p))(o2 (S p))
(o3 (S p))(pc (S p))(t1 (S p)).

```

3) *Time-Related Part*: First we need to attach a time to each state in the sequence. This is done by function *fc* that is monotonous.

Definition Time := nat.

Variable fc : nat -> Time.

Hypothesis fc_monotonic : forall p, fc p < fc (S p).

A TON-timer is modeled by three hypotheses that constitute the time related predicates part of the TSTS:

- **Reset**. If $pc = e_2$, which means the timer rung is executed, and the timer is not enabled, then the timer bit is false at next state.

```

Hypothesis h_t1_reset_c: forall p, pc p = e2 ->
 $\sim m_1 p \rightarrow \sim t_1(S p).$ 

```

- **Set**. Given two indexes p_1 and p_2 , where $pc p_1 = pc p_2 = e_2$. If the time interval between them is greater than or equal to the preset time of the timer and the timer is enabled during this period, then the timer bit is true at next position.

```

Hypothesis h_t1_set_c : forall p1 p2, pc p1 = e2 ->
pc p2 = e2 -> t1_PT_c <= fc p2 - fc p1 ->
being_true_c m1 p1 p2 e2 pc -> t1(S p2).

```

- **Set Reverse**. If the timer bit is true at position p_2 and the timer instruction was executed at the previous position, then there exists a position p_1 before p_2 such that the time interval between $pred p_1$ and p_2 is great than or equal to $t1_PT_c$ and during this period it is enabled.

```

Hypothesis h_t1_true_c: forall p2, pc(pred p2) = e2 ->
t1 p2 -> exists p1, pc p1 = e2 \ \wedge t1_PT_c <= fc(pred p2)
- fc p1 \ \wedge being_true_c m1 p1 (pred p2) e2 pc.

```

Finally we have the TSTS for the concrete model.

B. Outline of Refinement Proof

As mentioned in section II, we first need to define two functions: one is a state mapping ϕ from the concrete states to the abstract states and the other is a sampling mapping f of type $N \rightarrow N$. These two functions are used to construct an abstract sequence from a concrete one. We need to prove that the obtained abstract sequence is a timed execution sequence of the abstract model. In order to do so, it is sufficient to prove that the obtained abstract sequence satisfies the initiation condition, consecution condition and the timing relation of the abstract model.

Since the variables used in the two models are the same, ϕ is an identical function and is omitted.

1) *Defining f*: Given a natural n , $f n$ is the index of the state in the concrete sequence corresponding to the n -th state in the abstract sequence. f is built using the method proposed in [8] that utilizes a predicate which indicate whether the sampling operation can take place. P is the predicate that is true when the program counter is e_{11} . Intuitively, $f n$ denotes the index at which P is true for the n -th time. *abs* is a help function which maps a concrete sequence to an abstract one using the sampling function f .

Definition P n := pc n = e11.

```

Definition f_mapping_0_to_Prop (x:nat)(P:nat->Prop):=
(forall x', x' < x ->  $\sim P x'$ ) \ \wedge P x.

```

Definition f_mapping_0_to :=

```

epsilon nat_non_empty (f_mapping_0_to_Prop P).

```

Definition f_mapping_S_n_to_Prop (n x:nat)(P:nat->Prop)

```

:= (forall x', n < x' < x ->  $\sim P x'$ ) \ \wedge P x \ \wedge (n < x).

```

Definition f_mapping_S_n_to(n:nat):=

```

epsilon nat_non_empty(f_mapping_S_n_to_Prop n P).

```

Fixpoint f (n : nat) := match n with

```

0 => f_mapping_0_to

```

```

| S p => f_mapping_S_n_to (f p)

```

end.

Definition abs(T:Type)(v:Var_c)(n:nat):= v(f n).

2) *Stating the Assumptions and Constraints*: The assumptions and constraints described in [1] are formally stated. We take three of them for example:

- The executions of the same instruction take the same time.

Hypothesis ExeSameTime :

```

forall i1 i2 l, pc i1 = l \ \wedge pc i2 = l ->
f (S i1) - f i1 = f (S i2) - f i2.

```

- There is no loop in one scan cycle.

Hypothesis NoLoop :

```

forall i1 i2, P i1 \ \wedge P i2 -> forall i3 i4,
i1 < i3 < i2 \ \wedge i1 < i4 < i2 \ \wedge i3 <> i4 ->
pc i3 <> pc i4.

```

- Each relay can be set value at most once per scan cycle.

Hypothesis SetOnceM1 :

```

forall i1 i2, P i1 \ \wedge P i2 -> forall i3 i4,
i1 < i3 < i2 \ \wedge i1 < i4 < i2 \ \wedge
m1 (S i3) <> m1 i3 \ \wedge m1 (S i4) <> m1 i4 ->
i3 = i4.

```

3) *Initiation Condition*: Given a timed state sequence ζ_C of C , we need to prove that $\zeta_A = \phi \circ \zeta_C \circ f$ is a sequence satisfying $\Theta_A(proj_1(\zeta_A(0)))$. $proj_1$ is a function to get the first element of a pair. The theorem is stated as follow and *initial_condition_a* is the initiation condition of the abstract model.

Theorem concrete_sequence_sat_abs_init_condition:

```

initial_condition_a (abs i0 0)
(abs i1 0)(abs i2 0)(abs i3 0)(abs i4 0)(abs m1 0)
(abs m2 0)(abs m3 0)(abs m4 0)(abs m5 0)(abs o0 0)
(abs o1 0)(abs o2 0)(abs o3 0)(abs t1 0).

```

4) *Consecution Condition*: Given a timed state sequence ζ_C of C , we need to prove that $\zeta_A = \phi \circ \zeta_C \circ f$ is a sequence satisfying $\forall i \in N, \rho_A(proj_1(\zeta(i), proj_1(\zeta(i+1))))$. The theorem about the consecution is stated below. And *next_condition_a* is the consecution condition of the abstract model.

```

Theorem concrete_sequence_sat_abs_next_condition :
forall p, next_condition_a(abs i0 p)(abs i1 p)
(abs i2 p)(abs i3 p)(abs i4 p)(abs m1 p)(abs m2 p)
(abs m3 p)(abs m4 p)(abs m5 p)(abs o0 p)(abs o1 p)
(abs o2 p)(abs o3 p)(abs t1 p)(abs i0(S p))(abs i1
(S p))(abs i2(S p))(abs i3(S p))(abs i4(S p))
(abs m1(S p))(abs m2(S p))(abs m3(S p))(abs m4(S p))
(abs m5(S p))(abs o0(S p))(abs o1(S p))(abs o2(S p))
(abs o3(S p))(abs t1(S p)).

```

5) *Timing Relation*: This proof of this result is done based on the theorems describing constraints and time related predicates of $\|A\|$. If we prove the theorems directly on the concrete model, a lot of irrelevant information will be introduced which makes the proof much more difficult. The theorems about the constraints can help to remove the irrelevant information. In practice, they are useful. The following theorems are about the timed related predicates and are similar to those hypotheses of the abstract model except that each variable is replaced by its abstract version. They are proved based on the theorems of constraints and timed related predicates of the concrete model.

```

Theorem concrete_sequence_sat_abs_t1_reset :
forall c, ~abs m1 c->~abs t1
Theorem concrete_sequence_sat_abs_t1_true :
forall c2,abs t1 c2->exists c1,
t1_PT <= abs fc (pred c2)-abs fc(pred c1)
/\being_true (abs m1) c1 c2.
Theorem concrete_sequence_sat_abs_t1_set :
forall c1 c2, t1_PT <= abs fc(pred c2)-
abs fc(pred c1)->
being_true (abs m1) c1 c2 -> abs t1 c2.

```

6) *The Final Theorem*: The final result is that for all properties that are held by the abstract model are also held by the concrete model according to the sampling function. In the theorem P is a property of the abstract model and sequences of form “ $abs \bullet$ ” are the sampled sequences. By using this theorem, the properties we have proved in [1] are also satisfied by the concrete model – there is no need to prove them on the concrete model from scratch.

```

Theorem refinement_relation_property :
forall P (f:Cycle->Time)
(i0 i1 i2 i3 i4 m1 m2 m3 m4 m5 o0 o1 o2 o3 t1:Var),
(initial_condition_a(i0 0)(i1 0)(i2 0)(i3 0)(i4 0)
(m1 0)(m2 0)(m3 0)(m4 0)(m5 0)(o0 0)(o1 0)(o2 0)
(o3 0)(t1 0)) ->
(forall p : Cycle,next_condition_a (i0 p)(i1 p)
(i2 p)(i3 p)(i4 p)(m1 p)(m2 p)(m3 p)(m4 p)(m5 p)
(o0 p)(o1 p)(o2 p)(o3 p)(t1 p)(i0(S p))(i1(S p))
(i2(S p))(i3(S p))(i4(S p))(m1(S p))(m2(S p))
(m3(S p))(m4(S p))(m5(S p))(o0(S p))(o1(S p))
(o2(S p))(o3(S p))(t1(S p))) ->

```

```

(forall c,~m1 c->~t1 c)->(forall c1 c2,
t1_PT<=f(pred c2)-f(pred c1)->being_true m1 c1 c2
-> t1 c2) ->(forall c2,t1 c2->
exists c1,t1_PT<=f(pred c2)-f(pred c1)/\
being_true m1 c1 c2)->
P i0 i1 i2 i3 i4 m1 m2 m3 m4 m5 o0 o1 o2 o3 t1)->
P(abs i0)(abs i1)(abs i2)(abs i3)(abs i4)(abs m1)
(abs m2)(abs m3)(abs m4)(abs m5)(abs o0)(abs o1)
(abs o2)(abs o3)(abs t1).

```

7) *Remarks*: After analyze the dependence among theorems used in the proof, it can be concluded that the proof of the refinement relation between abstract and concrete time related predicates only uses the theorem of constraints and concrete timed related predicates, which means that the refinement relation between abstract and concrete models are independent of the STS of the concrete model if the concrete model satisfies the constraints. By modifying the source codes or generating codes that satisfy the constraints, the refinement relation between the time related parts is held without proving.

IV. CONCLUSIONS

The paper addressed the formal validation of PLC systems with timers in the theorem proving system Coq. The timer behavior is characterized formally. A refinement validation methodology is presented in terms of an abstract model and its interpretation. The interpretation was calibrated by a mapping relation. The soundness of the methodology is shown in the proving system. An illustrative case study demonstrates the effectiveness of the approach.

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