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Time-data tradeoff for the sparse and cosparse regularizations of physics-driven inverse problems

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Abstract—We investigate the computational performance of the sparse vs cosparse regularizations applied to physics-driven inverse problems, relative to the amount of measurements. Our results show that, despite nominal equivalence of the two models in the given context, the analysis-based optimization benefits from an increase in the volume of available data, while the synthesis one does not.

I. INTRODUCTION

In the context of linear inverse problems, the goal is to estimate a signal $\mathbf{x} \in \mathbb{R}^n$ given measurements $\mathbf{y} = \mathbf{M}\mathbf{x} + \varepsilon$, where \mathbf{M} is a linear observation operator and ε is additive noise. In most cases, the problem is ill-posed (e.g. $\mathbf{M} \in \mathbb{R}^{m \times n}$, $m < n$) and prior information is needed.

In many problems of interest, the first layer of prior information is driven by physics. Examples include acoustic pressure [1], [2], electric potentials in electroencephalography [3], [4] and nuclear magnetization in magnetic resonance imaging [5]. This type of information can be formulated in two ways. One formulation is through linear partial differential equations. In compact notation:

$$\mathcal{A}(X) = Z, \quad (1)$$

where \mathcal{A} is a differential operator (including the associated boundary conditions) affecting the *continuous* signal X on the entire domain of interest Ω . The right term Z usually represents “sources” or “sinks”. Another form of representation is through the superposition principle, or compactly:

$$\mathcal{D}(Z) = X, \quad (2)$$

where \mathcal{D} is an integral operator generated from all solutions $\mathcal{A}(X) = \delta_i, \forall i \in \Omega$. These solutions are known as *Green’s functions* [6].

The second layer of information is the sparsity assumption. Indeed, in many cases, the right hand side of (1) will be equal to zero, except at singularity points in Ω . We now consider discretized versions of the above vectors and operators: $X \rightarrow \mathbf{x}$, $Z \rightarrow \mathbf{z}$, $\mathcal{A} \rightarrow \mathbf{A}$ and $\mathcal{D} \rightarrow \mathbf{D}$. Accordingly, the vector $\mathbf{z} \in \mathbb{R}^n$ is very sparse.

One way to exploit this prior knowledge is to consider the sparse synthesis [7] data model:

$$\mathbf{z}^* = \arg \min_{\mathbf{z}} F_d(\mathbf{M}\mathbf{D}\mathbf{z} - \mathbf{y}) + F_r(\mathbf{z}) \text{ where } \mathbf{x}^* = \mathbf{D}\mathbf{z}^*. \quad (3)$$

Here \mathbf{D} represents the *dictionary* of Green’s functions, and F_d , F_r are the data-fidelity and (sparsity-promoting) regularizer, respectively.

The alternative solution is obtained by the sparse analysis (a.k.a. *cosparse* [8]) data model:

$$\mathbf{x}^* = \arg \min_{\mathbf{x}} F_d(\mathbf{M}\mathbf{x} - \mathbf{y}) + F_r(\mathbf{A}\mathbf{x}), \quad (4)$$

where \mathbf{A} is the *analysis operator*.

For properly chosen boundary conditions, equations (1) and (2) have unique solutions. Consequently, the dictionary and the operator are non-singular with $\mathbf{D} = \mathbf{A}^{-1}$, and the two approaches become

nominally equivalent [9]. However, they are **substantially different from a computational point of view**: for most discretizations, the matrix \mathbf{A} contains $\mathcal{O}(n)$ non-zero elements, whilst the matrix \mathbf{D} is usually fully dense, i.e. containing $\mathcal{O}(n^2)$ non-zeros.

II. SHOWCASE: SOUND SOURCE LOCALIZATION

As an example, consider localizing sound sources given microphone recordings of the acoustic pressure X .

The physical model is given by the acoustic wave equation:

$$\Delta X(\mathbf{r}, t) - \frac{1}{c^2} \frac{\partial^2 X(\mathbf{r}, t)}{\partial t^2} = \sum_j Z(\mathbf{r}, t) \delta(\mathbf{r} - \mathbf{r}_j) \delta(t - t_j), \quad (5)$$

where c is the speed of sound, Z represents the contribution of sound sources, while \mathbf{r} and t represent the space and time coordinates.

In the discrete setting, this corresponds to the system of difference equations $\mathbf{A}\mathbf{x} = \mathbf{z}$ (obtained, e.g. by the Finite Difference Time Domain - FDTD method [10]). The number of variables is $n = st$, where s and t are the numbers of points used to discretize space and time, respectively. The measurements are given by $\mathbf{y} = \mathbf{M}\mathbf{x}$, where the row-reduced identity matrix $\mathbf{M} \in \mathbb{R}^{mt \times st}$ models an acquisition system with m microphones randomly distributed in space.

Solving the regularized¹ inverse problems (3) and (4) yields estimates of \mathbf{x} and \mathbf{z} . The source locations can then be easily obtained from the support of the source term \mathbf{z} .

III. SIMULATIONS

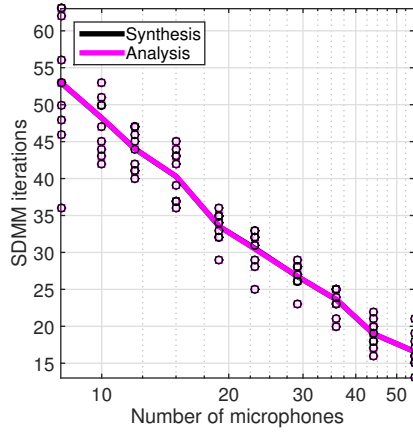
Some recent theoretical studies [12] suggest that the increase of training data enables acceleration of certain machine learning tasks. Inspired by this result, we investigate the behavior of the synthesis and analysis based sound source localization with regards to the amount of measurement data (i.e. the number of microphones m).

We simulate the acoustic wave field generated by 5 white noise sources in 2D domain. The wave field is in the form of a rectangular lattice of size $s = 15 \times 15$, propagating during $t = 50$ temporal instances. The number of microphones is varied, but such that all sources can be perfectly localized for all values of m .

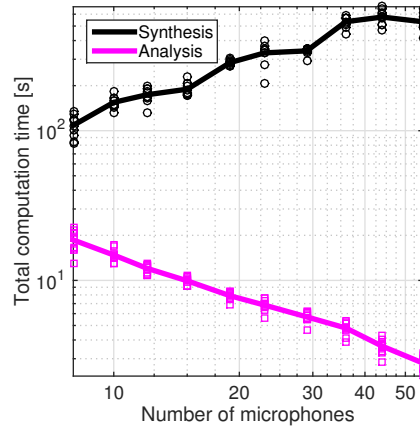
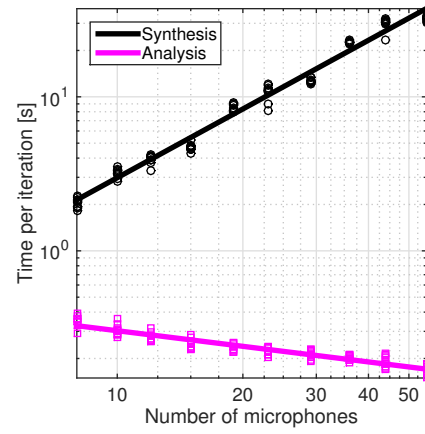
Figure 1(a) confirms that the convergence is indeed accelerated in terms of iteration count for the applied SDMM [13] algorithm. However, the cost per iteration grows for the synthesis-based optimization, as opposed to the analysis-based optimization (figure 1(b)). The overall result given in figure 1(c) is that only the latter benefits from raising the volume of measurement data. In the extreme case, the analysis version is two orders of magnitude faster.

The results are complementary to what has been concluded in [14], in the sense that the objective function is not changed. Hence, we envision further gains in computational performance by applying a smoothing strategy suggested in the article.

¹We used the hierarchical $\ell_{2,1}$ -norm [11] for F_r and $\ell_{\|\cdot\|_2 \leq \varepsilon}$ for F_d .



(a) Number of iterations vs number of microphones. (b) Processing time per iteration vs number of microphones.



(c) Total processing time vs number of microphones.

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